SOV/144-58-7-11/15

A Current Stabiliser with Non-linear Electro-magnetic Element time variations in current are not important.
There are 10 figures, 1 table and 4 Soviet references.

ASSOCIATION: Kafedra teorecicheskikh osnov elektrotekhniki Leningradskogo politekhnicheskogo institutu (Chair of Theoretical Fundamentals of Electrical Engineering, Leningrad Polytechnical Institute)

SUBMITTED: June 6, 1958

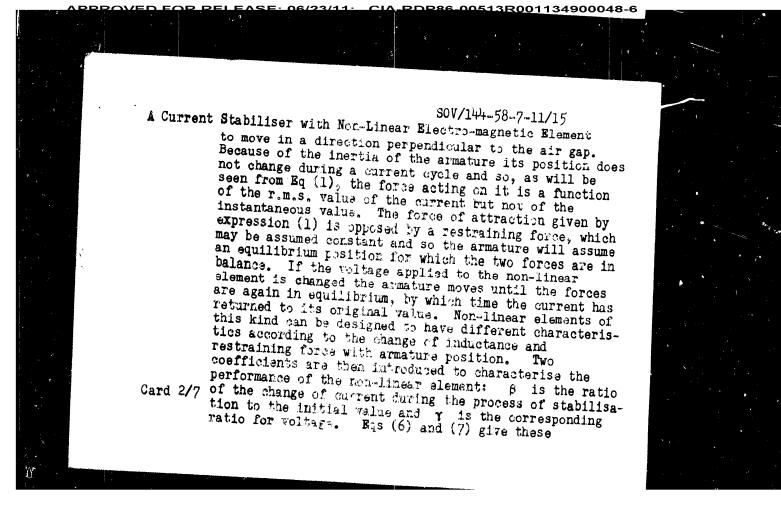
Card 7/7

SOV/144-58-7-11/15 A Current Stabiliser with Non-linear Electro-magnetic Element active resistance load. The vector diagram of thic circuit is given in Fig 9. Expression (16) is derived for the ratio of the power of the non-linear element to that of the receiving device and this equation is used to construct the family of curves of this ratio as a function of γ , (see Fig 10). These curves show that for a given ratio of maximum to minimum voltage on the circuit terminals there is a value of γ that it is best not to exceed. Test results are given in Table 1 for a circuit containing a non-linear element of power 5 kVAR; in this test the value of β was about 1%. seen from the table that in respect of weight per active power unit and in efficiency this current stabiliser is competitive with ferro-resonance voltage stabilisers. It is accordingly recommended to use this non-linear electro-magnetic element when inertia of the stabiliser is not an important defect. For example, it can be used to stabilise the filament current of large radio valves where because of the large thermal inertia short

307/144-58-7-11/15 A Current Stabiliser with Non-Linear Electro-Magnetic Element deriving the equations it was assumed that the magnetic reluctance of the leakage flux was constant; this is not so, but it does not have a serious effect on the value of γ . It does, however, affect β , for which the experimental value is always greater than the calculated. Therefore, the calculated value of \$\beta\$ does not always give sufficient information about the quality of current stabilisation. It is shown by a numerical example that when the gaps are small it is important to allow for the influence of the magnetic reluctance of the core on current stabilisation. It is then shown that the quality of current stabilisation can be improved if the armature is placed skew in the gap as shown in Fig 7. The effect of skewing the armature on the curve of current as a function of armature position will be seen from Fig 8. Skewing the armature is a fairly effective measure; for example, in one non-linear element the value of β was 2.3% and \gamma 140%. When the armature was skewed, for the same value of γ the value of β was only 1%. The case is then considered of a non-linear element used Card 5/7 as a current stabiliser in a circuit containing only

A Current Stabiliser with Non-Linear Electro-Magnetic Element placed in the air gap of the core. The leakage flux may be reduced and the coefficient γ increased by altering the arrangement of the coils as shown in Fig 5. With this type of non-litear element with small air gaps the coefficient γ is about 13% for a value of $\beta = 3.5%$. Besides being influenced by the coil arrangement the leakage flux also depends on the air gap geometry, particularly on the warie of length to depth (see Fig 5). The influence of this ratto on the performance of the device is then considered. In practice not all of the armature travel is available for current stabilisation and expression (15) gives the value of γ for a practical range of travel. It is then shown that the value of γ reaches a maximum when the ratio of the distance between the poles to the depth of the poles is about 0.5. Expression (15) was found to give values of γ in good agreement with experimental values. The accuracy of the calculation depends mainly on the accuracy of determination of the Laskage factor coefficient. It is more accurate to determine whis coefficient experimentally Gard 4/7 than to obtain it from the curves given in Fig 6.

SOV/144-58-7-11/15 A Current Stabiliser with Non-Linear Electro-Magnetic Element coefficients as functions of the parameters of the nonlinear element and analysis of these expressions shows that stabilisation is best when the magnetic reluctance of the core is small in comparison with the magnetic reluctance of the air gap between the poles; the coefficient β depends on the voltage range in which stabilisation is effected and the maximum value of the coefficient γ depends only on the ratio of the air gap between the poles when the armature is not in place to that when it is in place. The first type of non-linear element was made in accordance with the diagram shown in Fig 1 and was found to effect current stabilisation over only a limited range of voltage. The characteristic curve of this non-linear element is given in Fig 4, curve 1, from which it will be seen that stabilisation is effective only from the starting voltage of 90 V to about 130 V, above which stabilisation is much impaired. It is shown that the performance is not very satisfactory because the coefficient γ is small and this results Card 3/7 from the presence of large leakage fluxes and also from distortion of the field that coours when the armature is



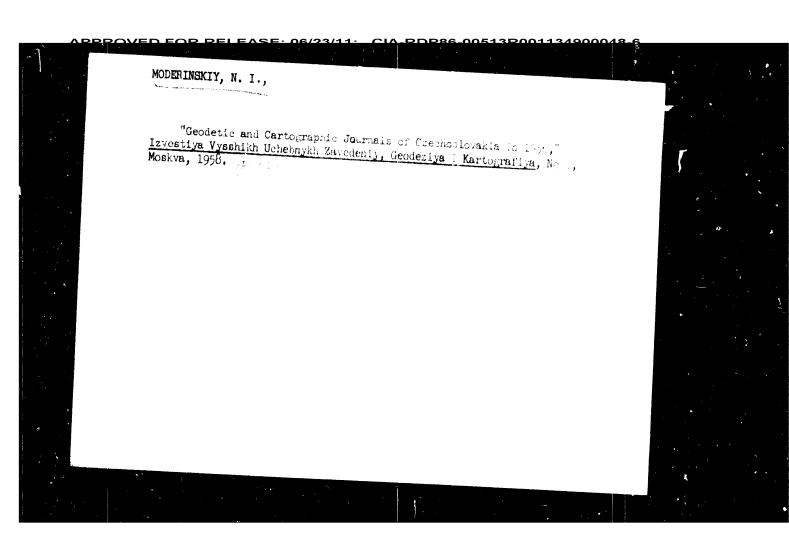
AUTHOR: Moderov, Andrey Aleksandrovich, Aspirant SOV/144-58-7-11/15 TITLE: A Current Stabiliser with Non-Linear Electro-magnetic Element (Stabilizator toka s nelineynym elektromagnitnym PERIODICAL: Izvestiya vysshikh uchebnykh zavedeniy Elektromekhanika, 1958, Nr 7, pp 99-111 (USSR) ABSTRACT: Many electrical devices supplied by a.c. employ a nonlinear relationship between r.m.s. values of voltage and current in some circuit components. Such stabilisers have the disadvantage of distorting current and voltage wave shapes. This defect is not shared by inertia nonlinear elements, the parameters of which do not change during a current cycle but depend on the r.m.s. value of the current. Stabilisers with inertia non-linear elements are also unaffected by changes in the frequency of supply. The electro-magnetic non-linear elements described below are of the inertia type and may be used for current or voltage stabilisation. stabiliser with non-linear electro-magnetic element is illustrated diagrammatically in Fig 1. It consists of a Card 1/7 reactive coil with an open core of ferro-magnetic material between the poles of which an armature is free

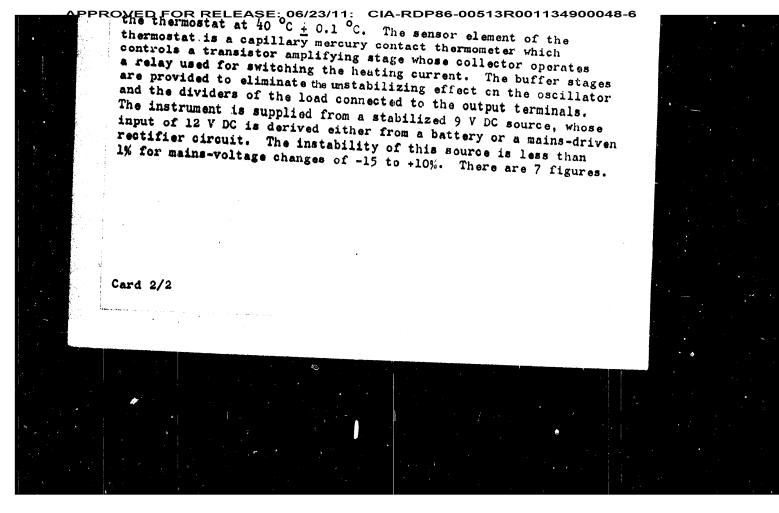
ICHICARLY, B., prof.; SAVIUK, V., insh. (Krayova, Emmynlya); GEART, F., hand, tekhn. muk (Budapenh, Vengriya); GERMETAKOV, V.M., insh.; M.OBROV, A.cha. insh.; SAPCHIKOV, R.A., doktor tekhn. neuk, prof.; tekhn. muk (Noscow).

Modification of the Heaviside formula. Hicktrichestvo no.3:86-88 Kr '50. (MEA 11:5)

1. Lodsinskiy politekhnicheskiy institut, Pol'sha (for Konorskiy). 2. Leningradskiy politekhnicheskiy institut iseni Lelinina (for Greeknyakov, Modorov). 3. Leningradskiy voyenno-sekhanicheskiy institut (for Saposhnikov, Sapershteyn).

(Hectric engineering)



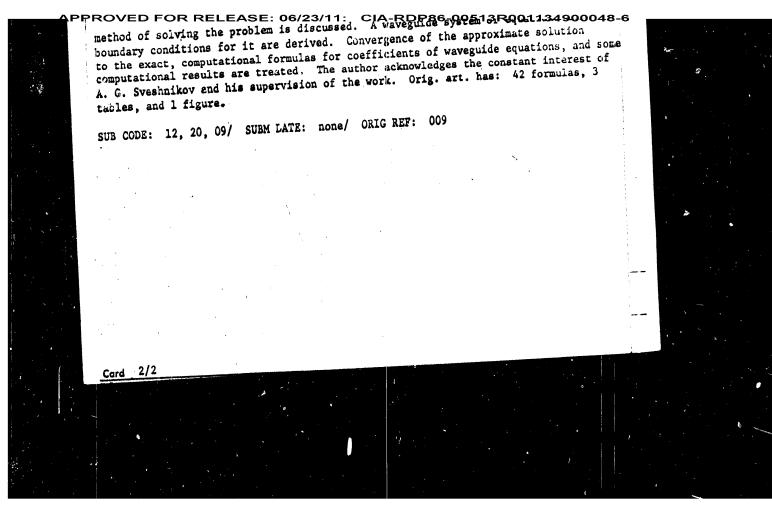


AUTHORS: Moder, Antonín and Sajal, Pavel, Engineers

TITLE: Frequency standard, type OTP

PERIODICAL: Sdělovací technika, no. 3, 1963, 93 - 94

The instrument consists of the following constructional TEXT: units: an oscillator operating at 100 kc/s; a thermal-control system; a frequency divider; buffer stages for frequencies of 100 kc/s, 20 kc/s, 5 kc/s and 1 kc/s and a stabilized DC supply source. The oscillator is based on a piezoelectric crystal unit of GT cut, which operates as a series resonant circuit. The crystal unit is sealed in an evacuated glass envelope provided with an octal base. The active element of the oscillator is a Czechoslovak transistor, type 156NU70. The oscillator is situated in a thermostat together with an amplifying stage inductively coupled to the oscillator. The supply for the oscillator is provided by a mercury battery, type MR 19, also situated in the thermostat. When it is necessary to change the battery the oscillator is connected to a stabilizer with a zener diode fed from a 9 V DC supply. The thermal-control system maintains the temperature of Card 1/2



ACC NR: AT6035245

SOURCE CODE: UR/3043/66/000/005/0197/0209

AUTHOR: Modenov, V. P.

ORG: none

TITLE: The design of irregular waveguides with gyrotropic filling

SOURCE: Moscow. Universitet. Vychislitel nyy tsentr. Sbornik rabot, no. 5, 1966.

Vychislitel'nyye metody i programmirovaniye (Computing methods and programming),

TOPIC TAGS: waveguide, dans irregular waveguide, gyrotropic body, approximate

ABSTRACT: The basic ideas of the method of designing waveguides with local gyrotropic filling have already been set forth. In particular, the method permits study of the phenomenon of electromagnetic wave reflection from a gyrotropic body of revolution of arbitrary form located axisymmetrically in a cylindrical waveguide of circular cross section (waveguide phase-shifter). The present paper aims at generalizing the method to the case of gyrotropically filled waveguides with irregular lateral surface. The general system of designing these waveguides which it sets forth makes it possible, for example, to solve the important problem of matching two round waveguides of differing diameter by means of a symmetrical adapter. The mutual influence of two irregularities (the lateral surface of the waveguide and its filling) may be studied. The design circuit may be successfully realized on modern high-speed computers. The 1/2

solution of the given problem is to change it from a boundary value problem for equations in partial derivatives to one for a finite system of ordinary differential equations (to the waveguide system). The method of differential chaning was used an equations (to the waveguide system). The method for the vaveguide system of the computer to solve the boundary value problem for the vaveguide system. The computer to solve the boundary value problem for the vaveguide system of this method is determined by order of the finite system, accuracy of integration, of this method is determined by order of the finite system, accuracy with which its coefficients are found. The authors express their graticules to A. G. Sveshnikov for his consideration in the work. Orig. art. has: 32 to tude to A. G. Sveshnikov for his consideration in the work. Orig. art. has: 32 to tude to A. G. Sveshnikov for his consideration in the work. Orig. art. has: 32 to tude to A. G. Sveshnikov for his consideration in the work. Orig. art. has: 32 to tude to A. G. Sveshnikov for his consideration in the work. Orig. art. has: 32 to tude to A. G. Sveshnikov for his consideration in the work. Orig. art. has: 32 to tude to A. G. Sveshnikov for his consideration in the work. Orig. art. has: 32 to tude to A. G. Sveshnikov for his consideration in the work. Orig. art. has: 32 to tude to A. G. Sveshnikov for his consideration in the work. Orig. art. has: 32 to tude to A. G. Sveshnikov for his consideration in the work. Orig. art. has: 32 to tude to A. G. Sveshnikov for his consideration in the work. Orig. art. has: 32 to tude to A. G. Sveshnikov for his consideration in the work. Orig. art. has: 32 to tude to A. G. Sveshnikov for his consideration in the work. Orig. art. has: 32 to tude to A. G. Sveshnikov for his consideration in the work. Orig. art. has: 32 to tude to A. G. Sveshnikov for his consideration in the work. Orig. art. has: 32 to tude to A. G. Sveshnikov for his consideration in the work. Orig. art. has: 32 to tude to A. G. Sveshnikov for his co

ACC NR: AP6025924

SOURCE CODE: UR/020004/006/004/0706/0713

AUTHOR: Modenov, V. P. (Moscow); Kalinina, L. I. (Moscow)

ORG: none

TITLE: Design of a circular waveguide with variable anisotropic occupation

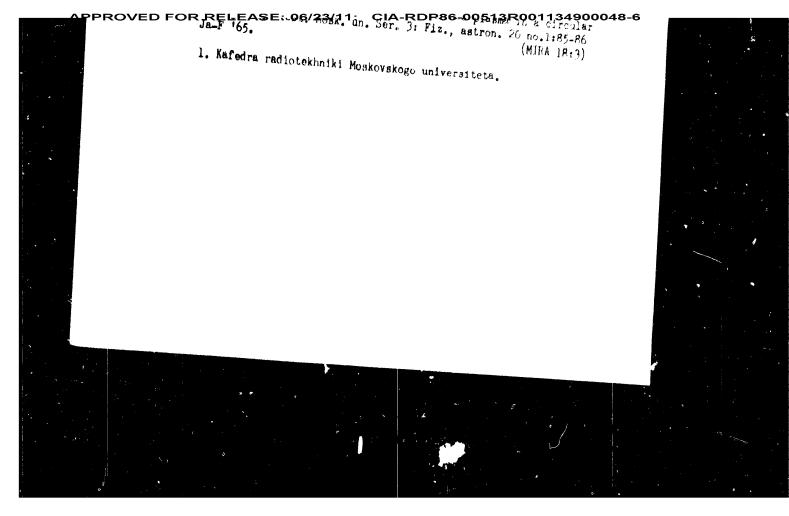
SOURCE: Zhurnal vychislitel'noy matematiki i matematicheskoy fiziki, v. 6, no. 4, 706-713

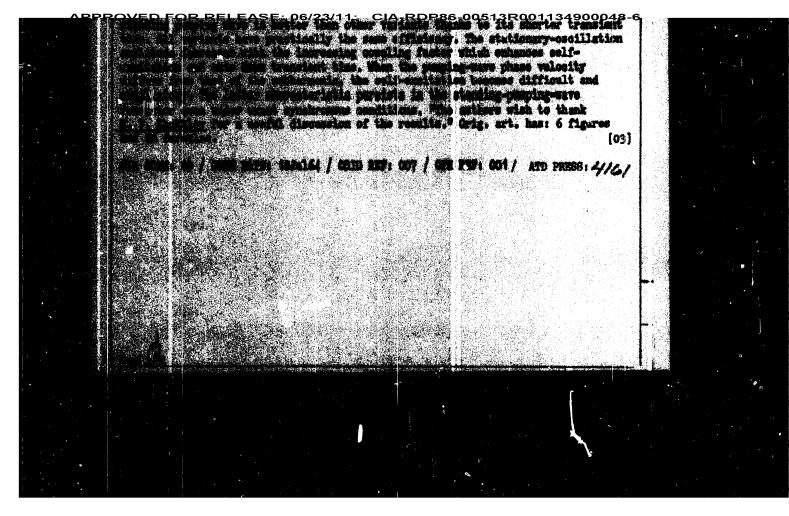
TOPIC TAGS: circular waveguide, differential equation solution, Maxwell equation, approximate solution

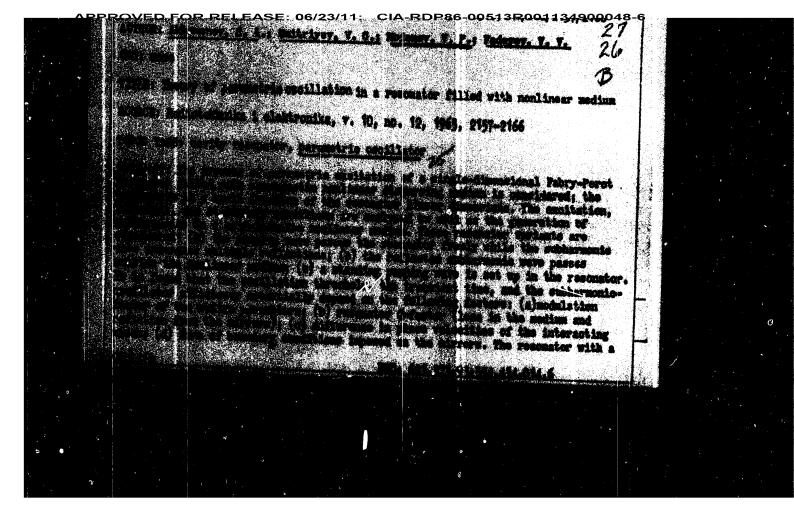
ABSTRACT: The authors attempt a methodical analysis and explanation of the nature of the computational error occurring when solving the initial waveguide problem by computer. The mathematical statement of the problem consists in determining in an irregular sector 0 < z < d of a variably and anisotropically occupied waveguide the solution of a Maxwell system of equations which must satisfy (1) the boundary condition of equality to zero of the tangential component of the strength of an electrical field on a lateral surface of the waveguide, (2) the conditions of conjugacy comprising continuos tangential components of the electrical and magnetic fields on the boundaries of the anisotropically occupied sector, and (3) the conditions of emission and excitation in regular sectors when there are no perpendicular waves coming from infinity except incident ones. The basic algorithm for constructing the approximate

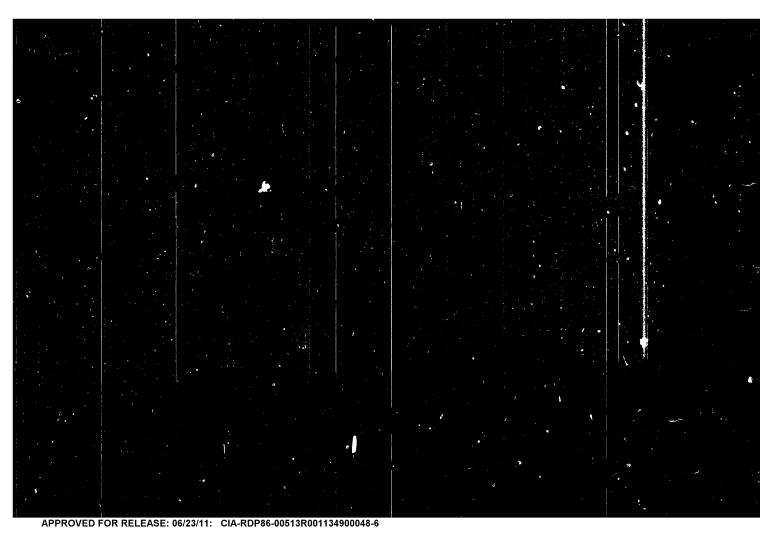
1/2

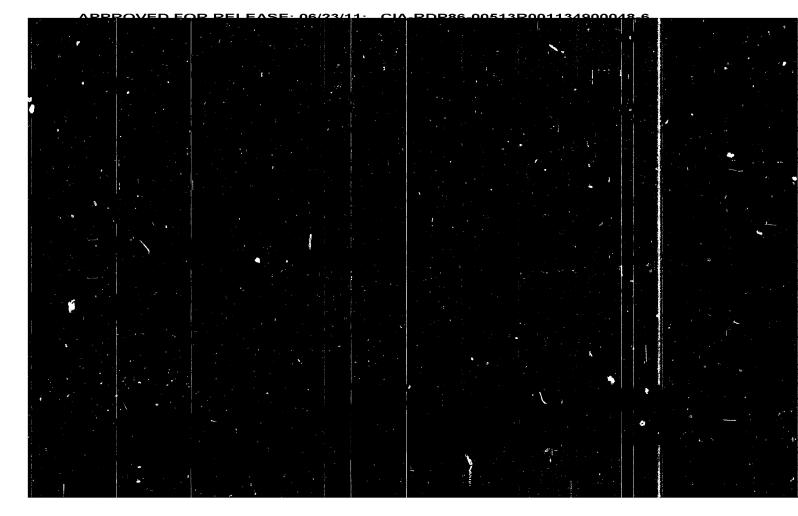
UDC: 517.9:621.372.8

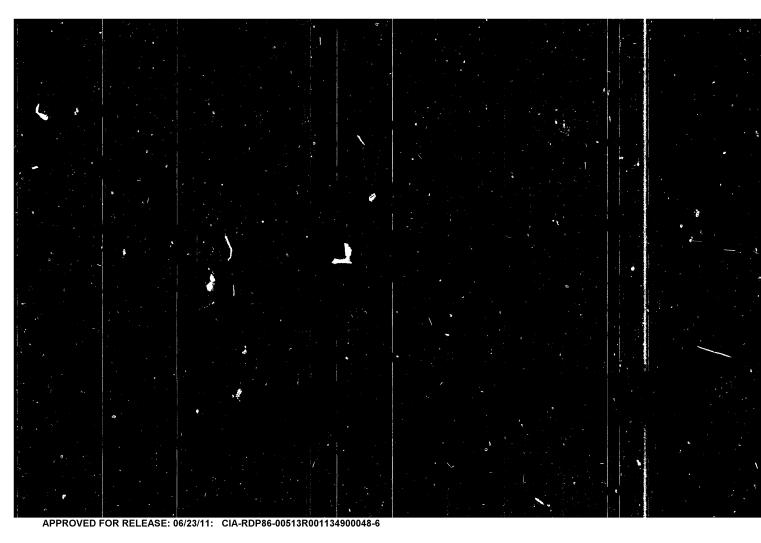


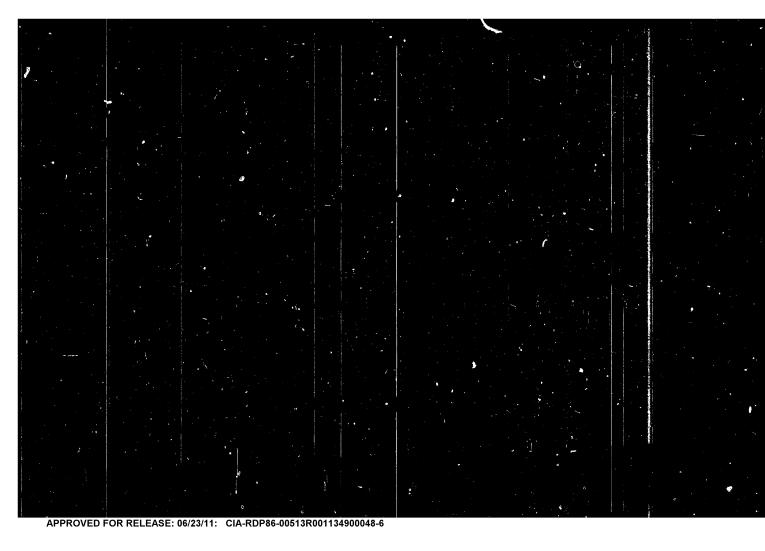


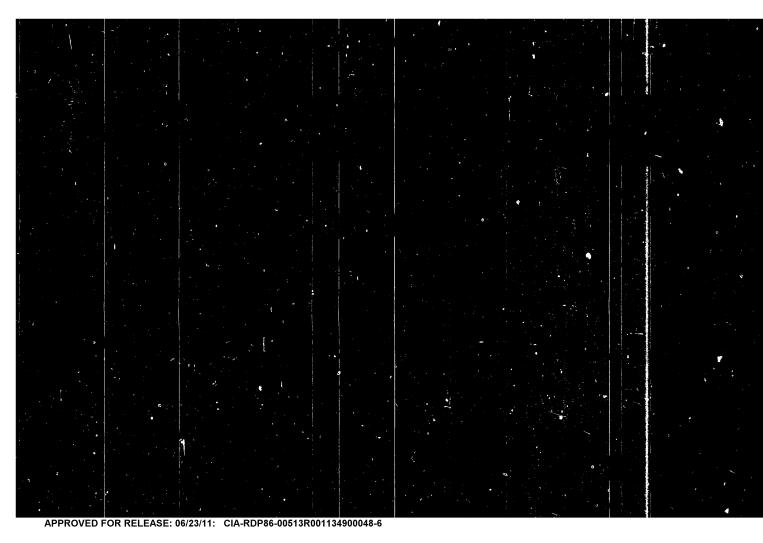


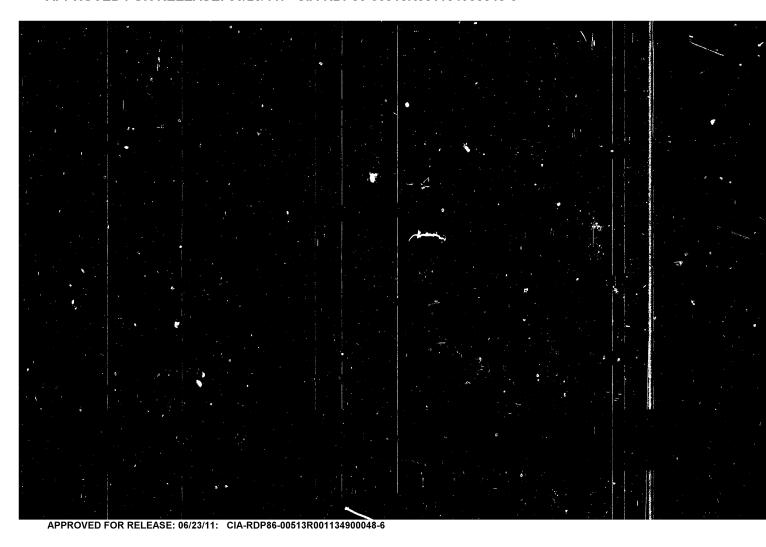












Ilinear dispersive line is practically possible; the tripling conversion factor may to keep down the conversion-equipment size. Orig. art. has: 4 figures and

ASSOCIATION: Moskovskiy gosudarstvenny*y universitet im. M. V.

SUBMITTED: 19Mar63 DATE ACQ: 05Jun64 ENGL: 00

SUB CODE: EC, NP NO REF SOV: 005 OTHER: 002

AUTHOR: Akhmanov, S. A.; Dmitriyev, V. G.; Modenov, V. P.

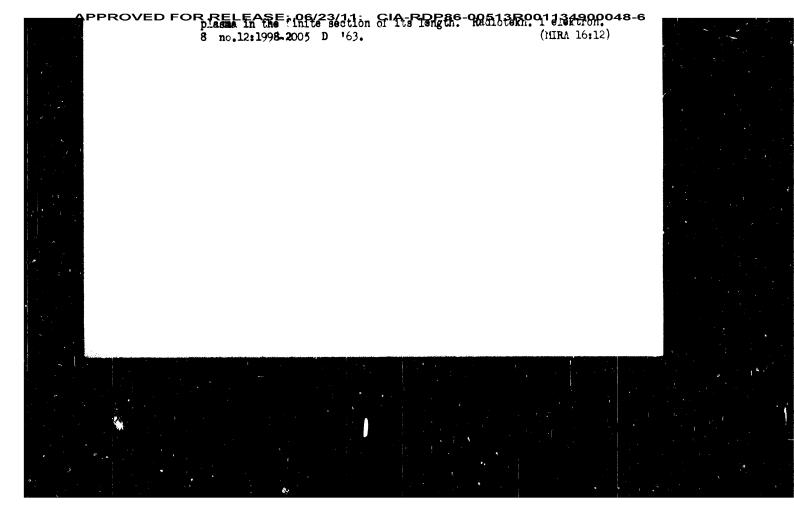
TITLE: Theory of frequency multiplication in nonlinear dispersive lines

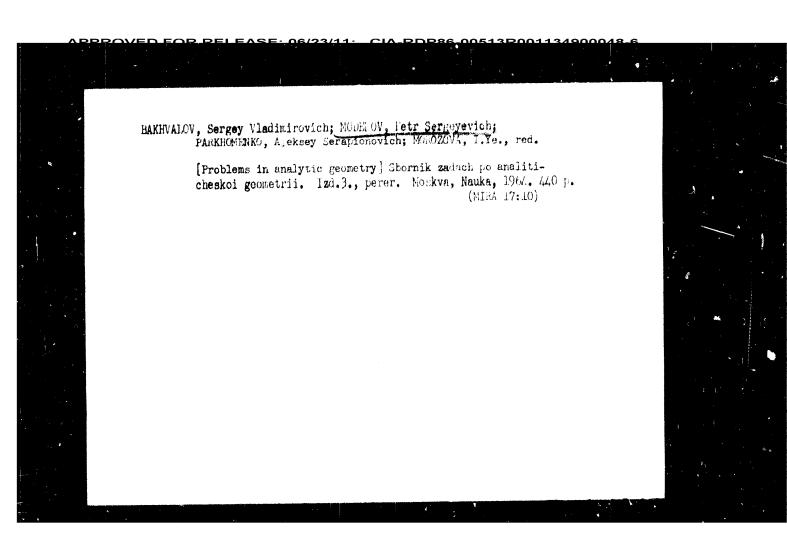
SOURCE: Radiotekhnika i elektronika, v. 9, no. 5, 1964, 814-821

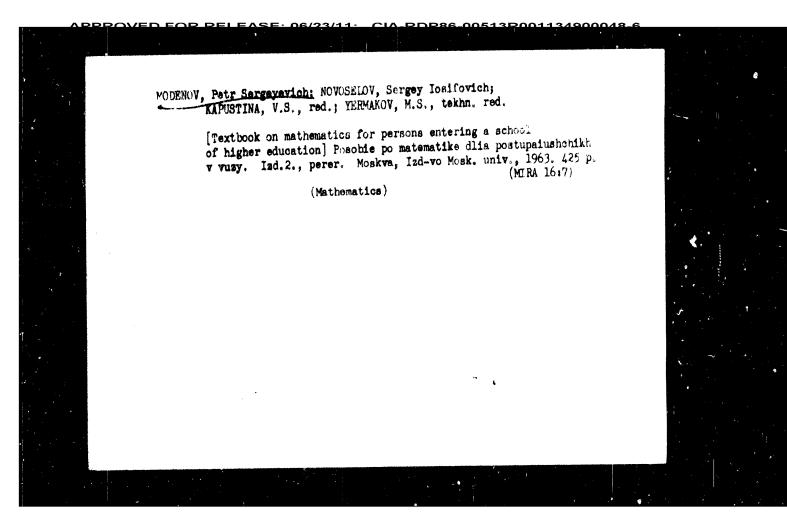
TOPIC TAGS: frequency multiplication, dispersive line, radio frequency multiplication, nonlinear optics

ABSTRACT: A theoretical investigation of the propagation of electromagnetic (radio and optical) waves in a nonlinear-reactance single-dimensional medium is reported; phase velocities of the fundamental wave and its second and third harmonics are regarded as nearly equal. The results may easily be extended over the case of a two-dimensional medium. The differential equations involved were numerically integrated on a "Strela" digital computer; the effects of the modulation factor, dispersion, and attenuation on the generation of harmonics

Core 1/2





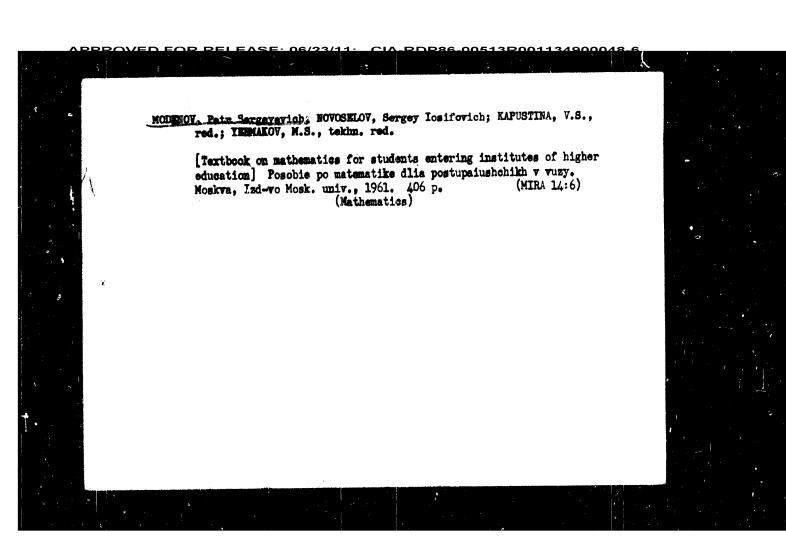


MODENCY, P.S. (Moskva)

Review of the journal "Journal de mathematique elementaire."

Mat. v shkole no.2:77-84 Mr-Ap '61. (MIRA 14:4)

(France--Mathematics--Periodicals)



MODEROV, Petr Sergeyevich; PARKECMENKO, Aleksey Serapionovich; KAPUSTINA, V.S., red.; YERMAKOV, M.S., tekhm. red. [Geometrical transformations] Geometricheskie preobrazovaniia.

Moskva, Izd-vo Mosk. univ., 1961. 230 p. (MIRA 15:2)

(Geometry, Projective) (MIRA 15:2) MODEROW, Petr Sergorovich: MOTORELOY, S.I., red.; SELAROYA, Ye.A., red.; SCHAROYS, S.S., tekhn.red.

[Collection of problems in a special course of elementary mathematical Scornik sadach po spetsial noun kursu elementarnoi metematiki. Ind.2., dop. 1 inpr. Monkva, Cos.ind-vo "Tyashaia shkola," 1960. 766 p. (MIRA 14:4)

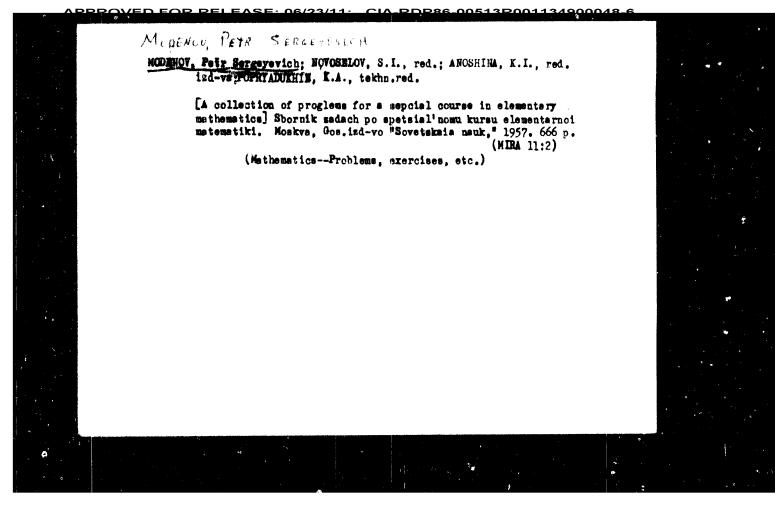
(Mathematics—Problems, exercises, etc.)

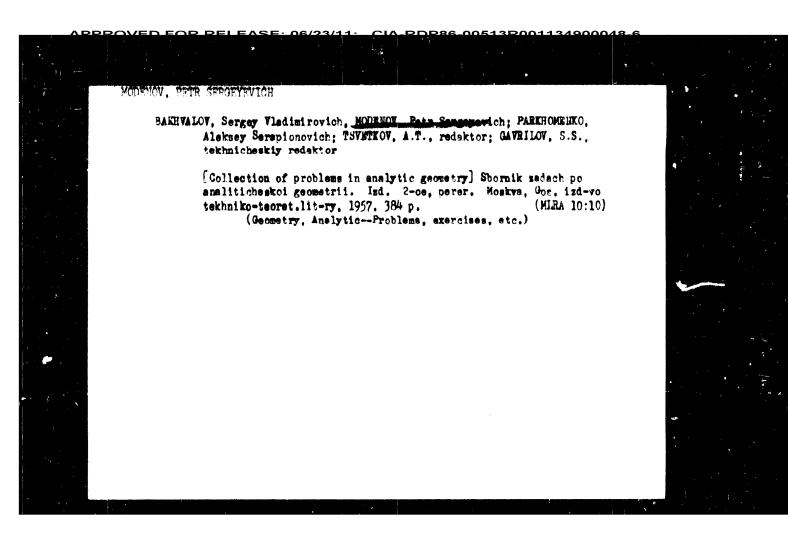
MORNOY, P.S. (Moscow); PARKINGENTO, A.S. (Moscow)

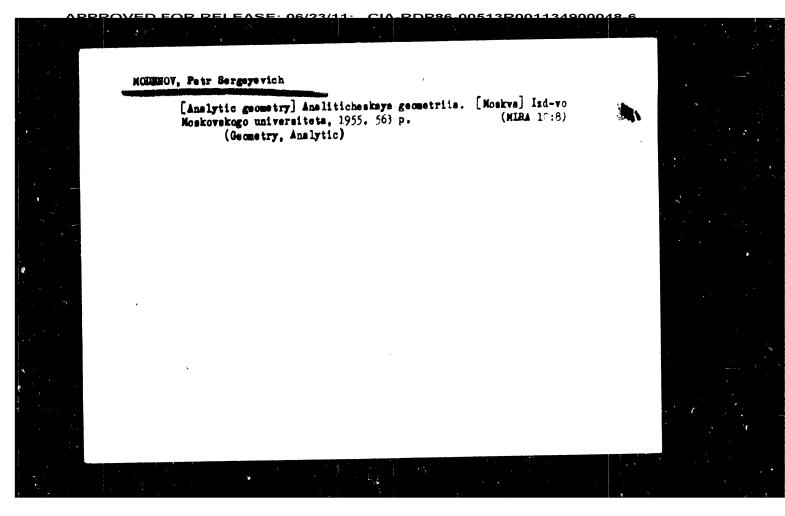
Projective plane and its topology. Mat. v shkole no. 4:5-17
Jl-Ag 158.

(Geometry, Projective)

(Topology)

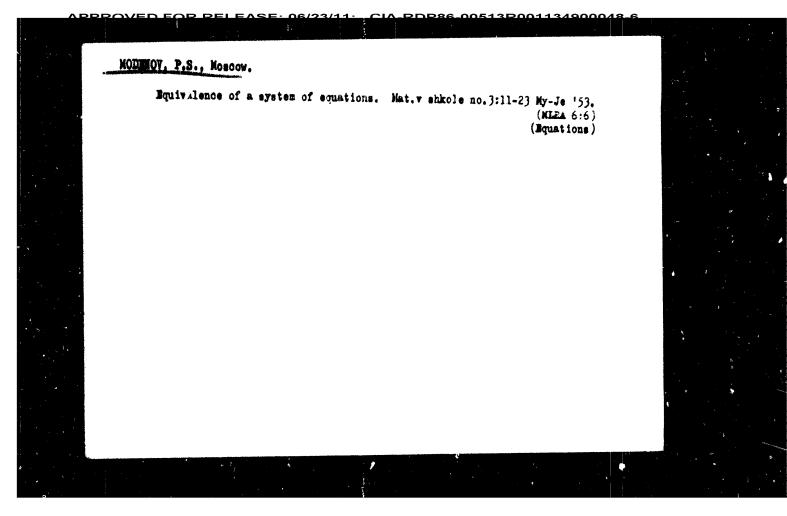




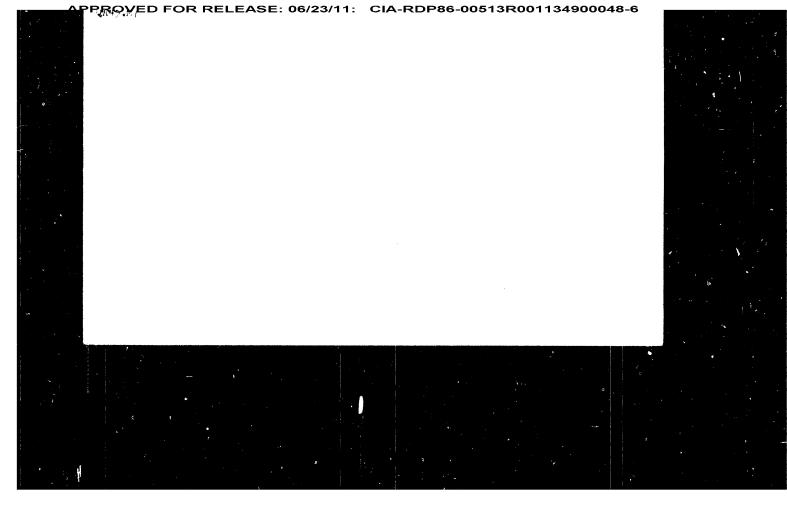


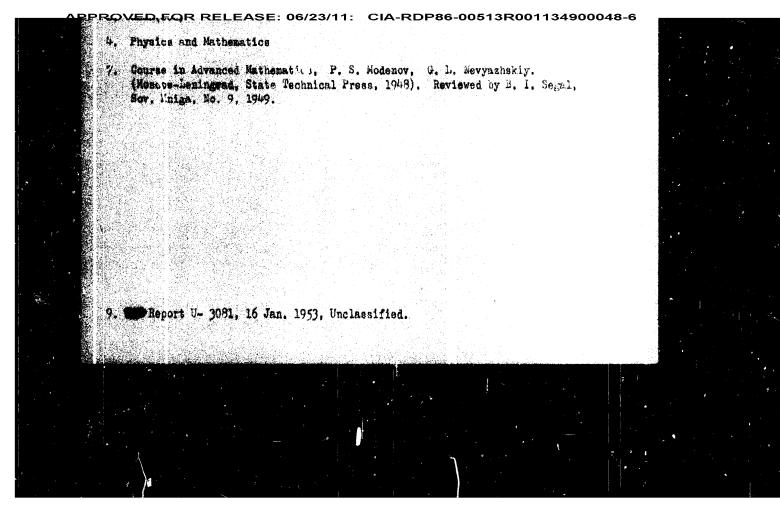
MODEROV, P. S. Metodicheskiye ukasani ya k programme po kureu "analiticheskaya geometriya." Diya studentov-sacchnikov i kurea fiz.matem. fak. ped. in-tov. m. uchpedgiz, 1954. 110 s. s. chert. 20am. (Olav. upr. podgotovki uchiteley m-va presveshchentya refer. nauch.-metod. kabinet po sacch. Obushentyu uchitey). 10,000 MZ. 1r. 25k.- (55-4108) 516 (071.4)

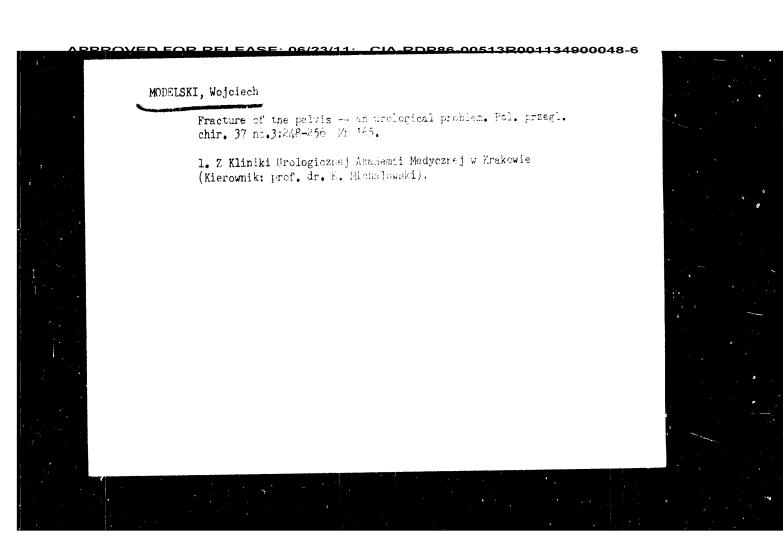
So. Knishnava Letopis', Vol. 7, 1955

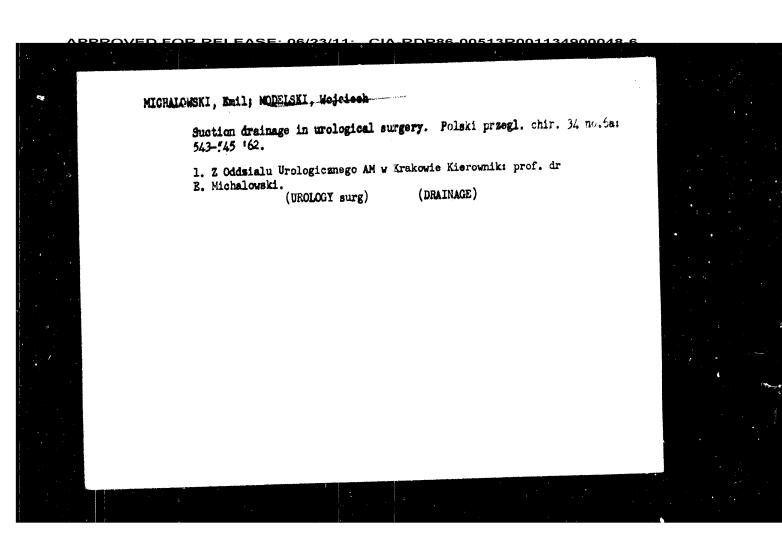


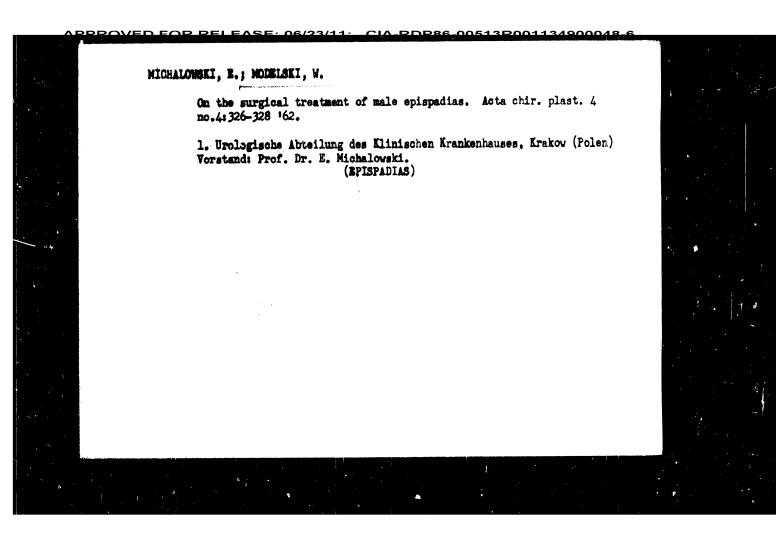
MODENOV, P. S. Froblems in mathematics Sbornik zadach po metematike. Izi. 4-3. Moskva, Sovetskaia hauka, 1953. 384 p. SO: Monthly List of Russian Accessions, Vol 6 No 6 September 1953

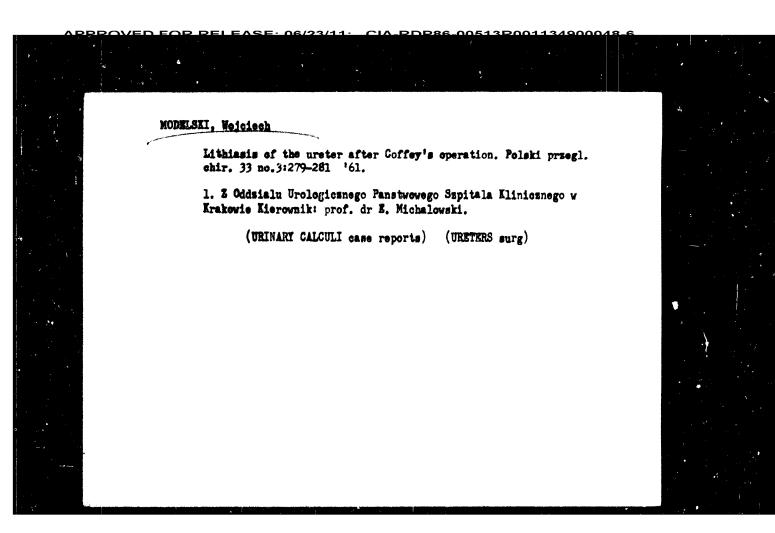


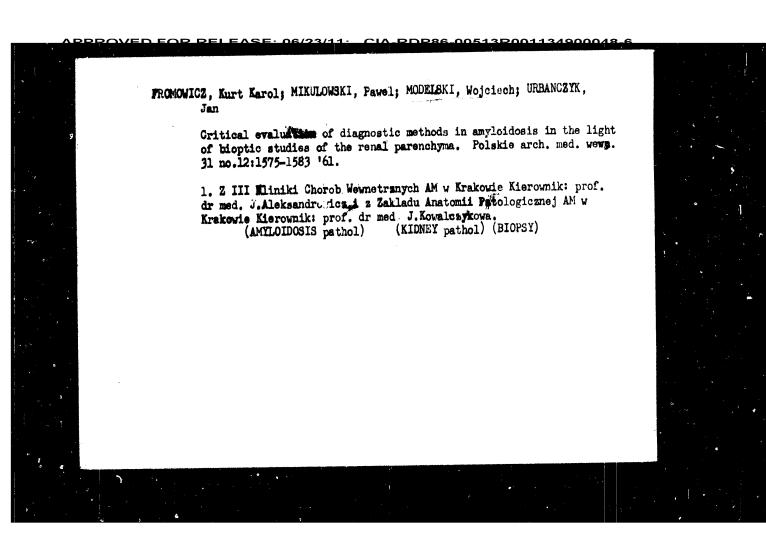


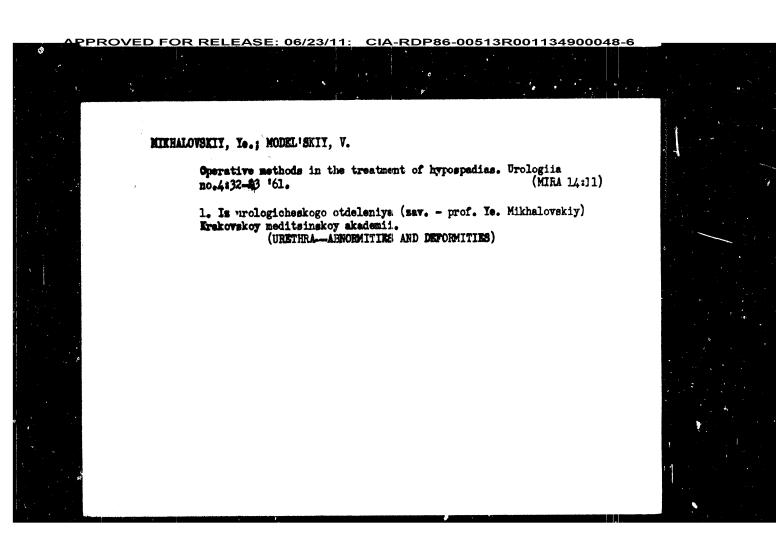


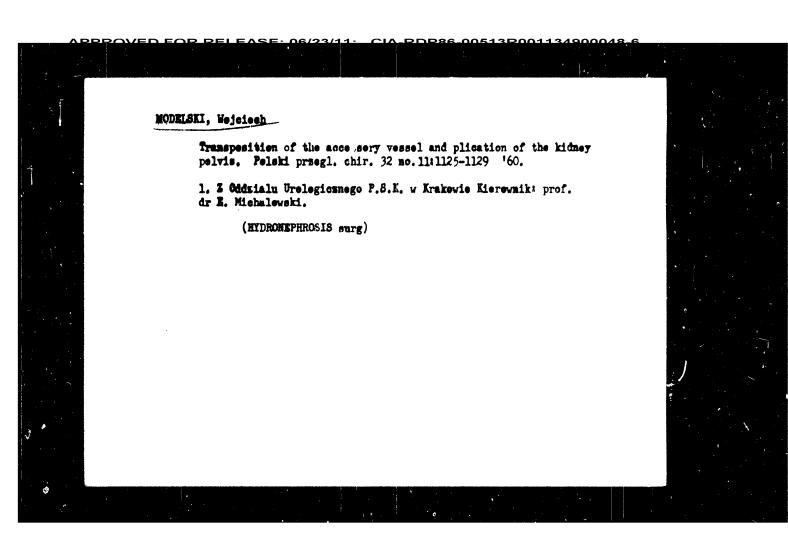


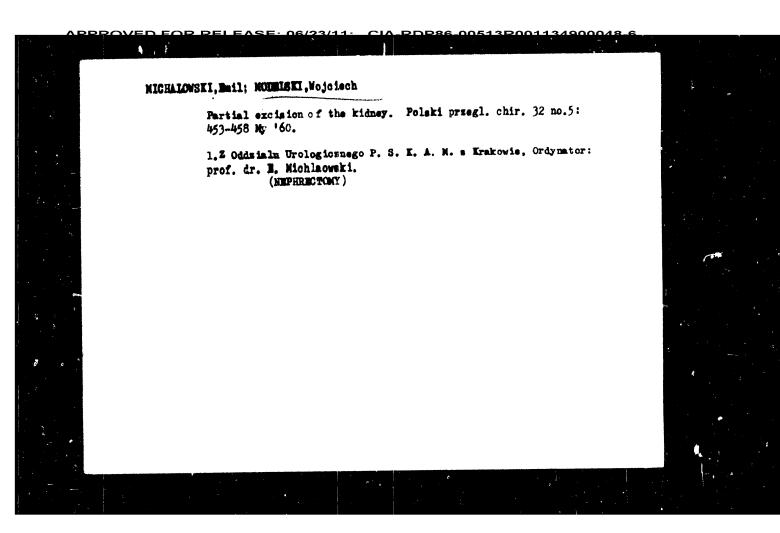


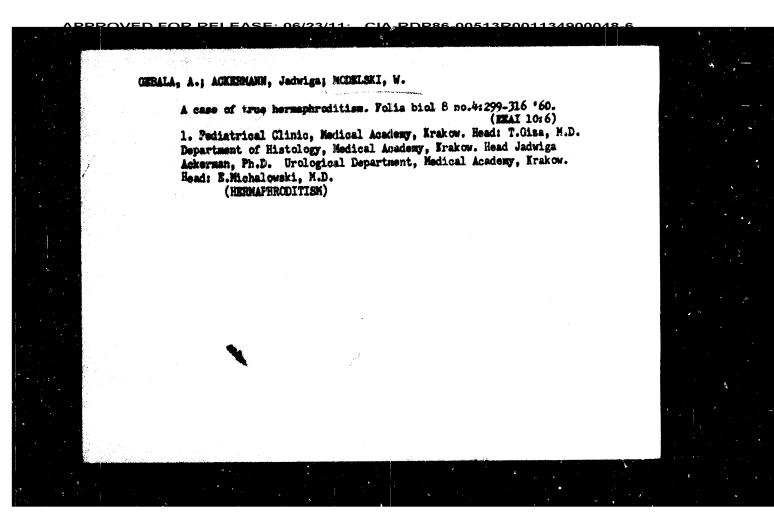


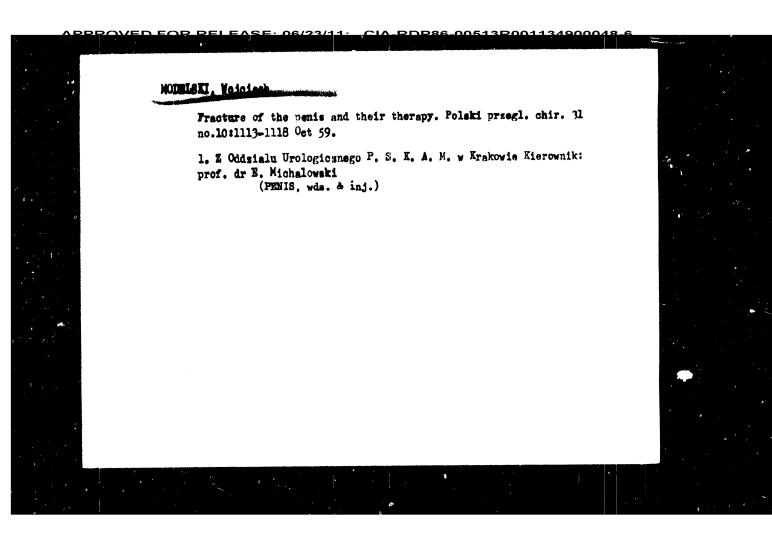


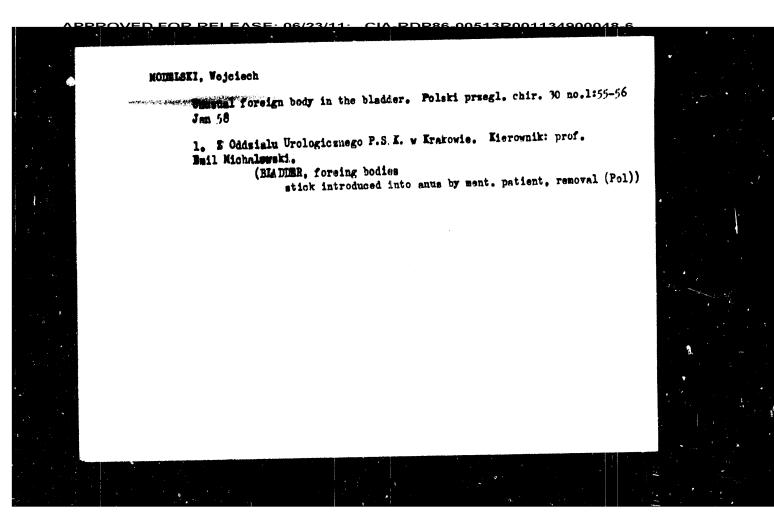








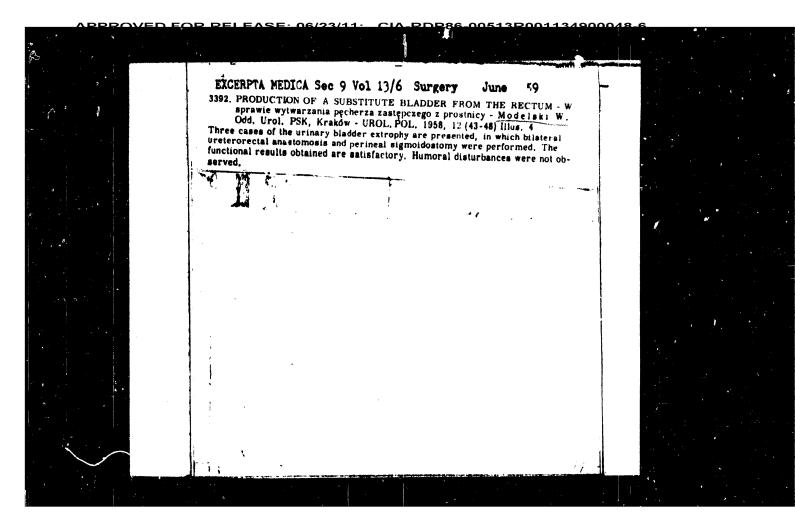


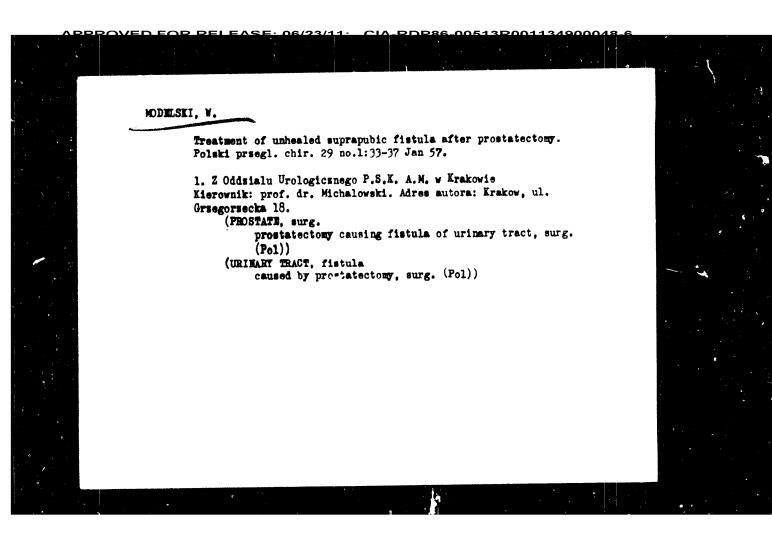


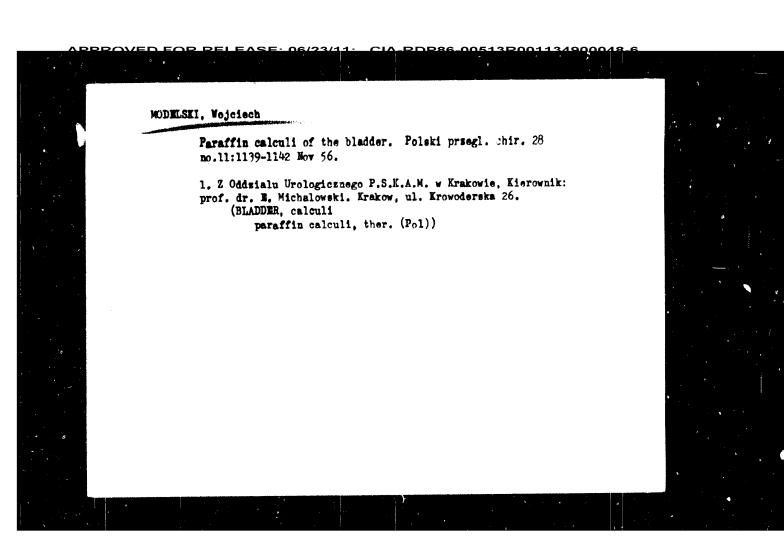
MODELETI

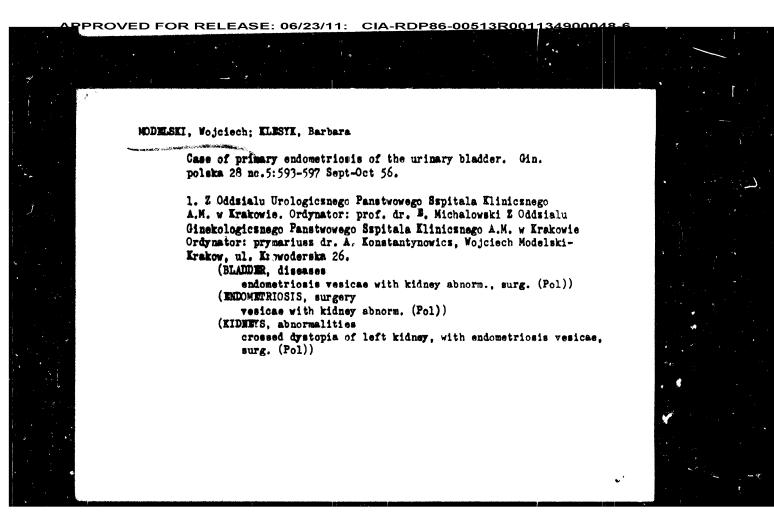
Diagnostic renal puncture in cases of hydronephrosis. Polski przegl.
radicl, 22 no.4:241-243 July-Aug 58.

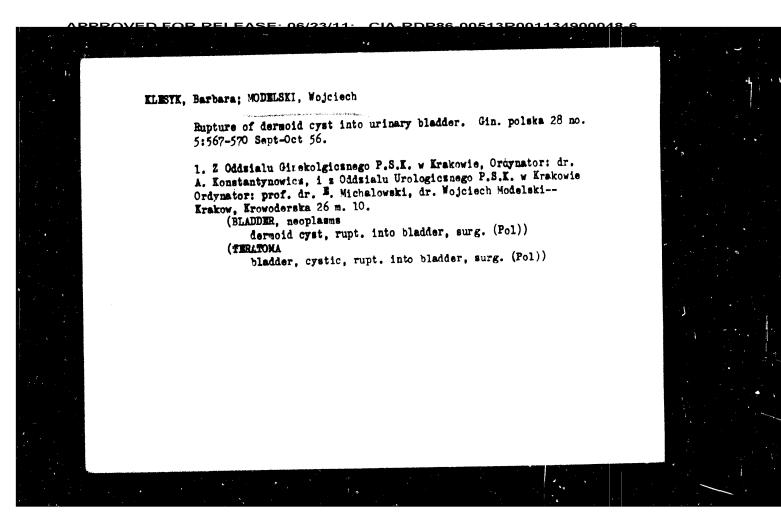
1. Z Oddsialu Urologiosnego P. S. K. v Krakowie Ordynator prof. dr.
E. Micholowski.
(HTIRONEPHROSIS, diag.
renal puncture, indic. & case reports (Pol))

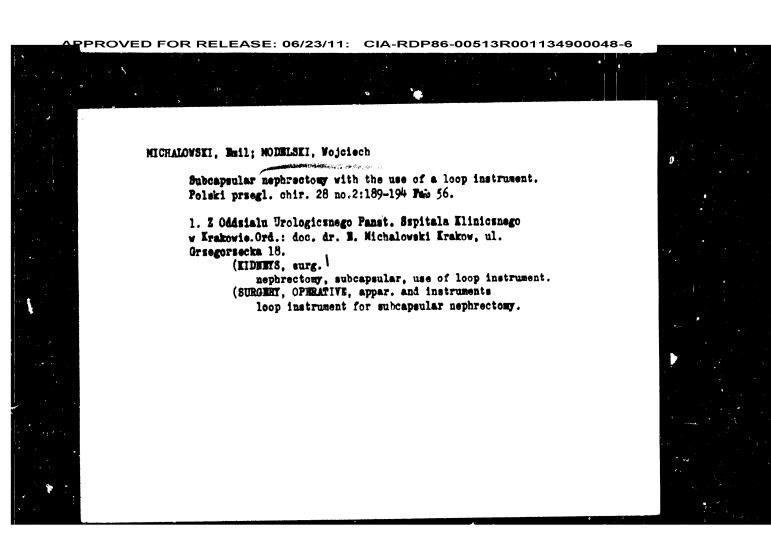


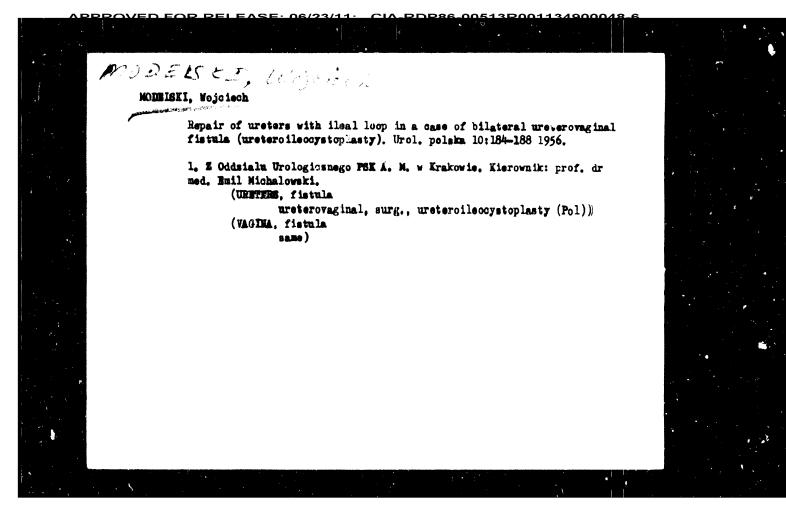












MICHALOWSKI, B.; NODELSKI, V.

Operative therapy of hydromephrosis due to obstruction of urstaro-pelvic junction. Postegy chir. 3:85-108 1956.

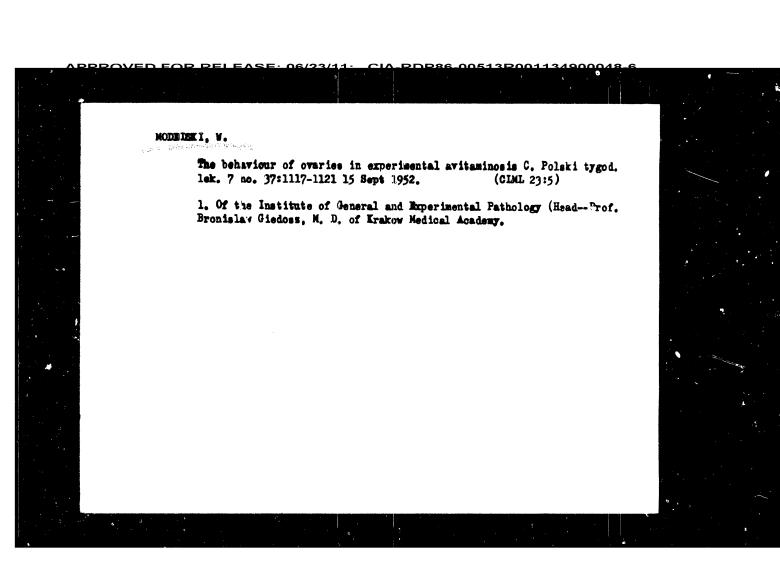
1. Z Oddsialu Urologicsnego Sspitala A.W. v Krakovis Ordynator prof. dr. sed. R. Michalowski.

(INTERNETIONIS, etc.). d pathogen.

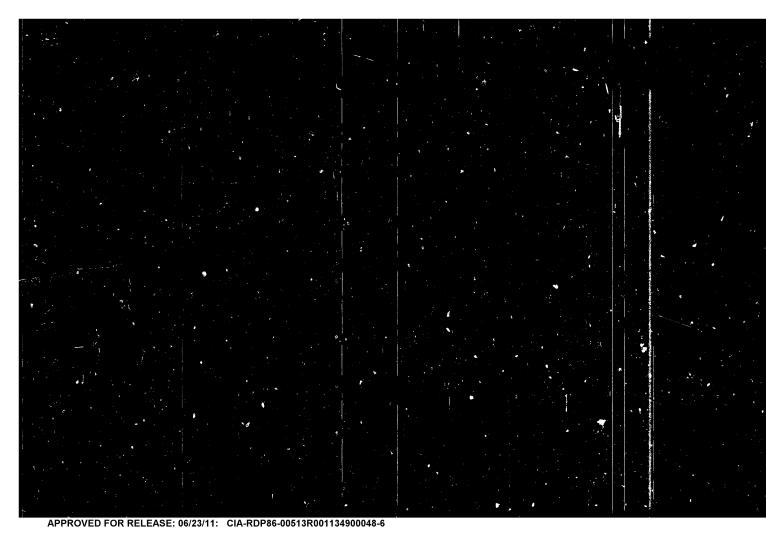
obstruct. of uretero-pelvic junction, surg., technics (Poll))

(URITHES, dis.
obstruct. of uretero-pelvic junction causing hydromephrosis, surg. 4 technics (Pol))

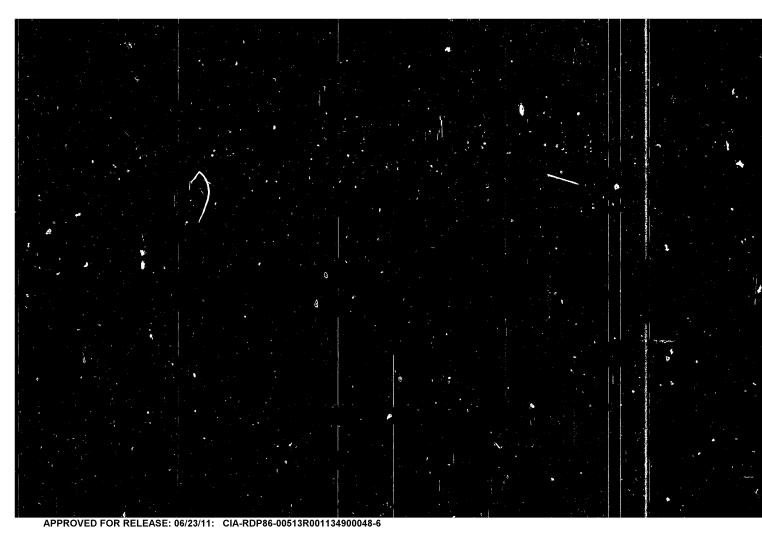
(KIDHE PEVIS, dis.
same)

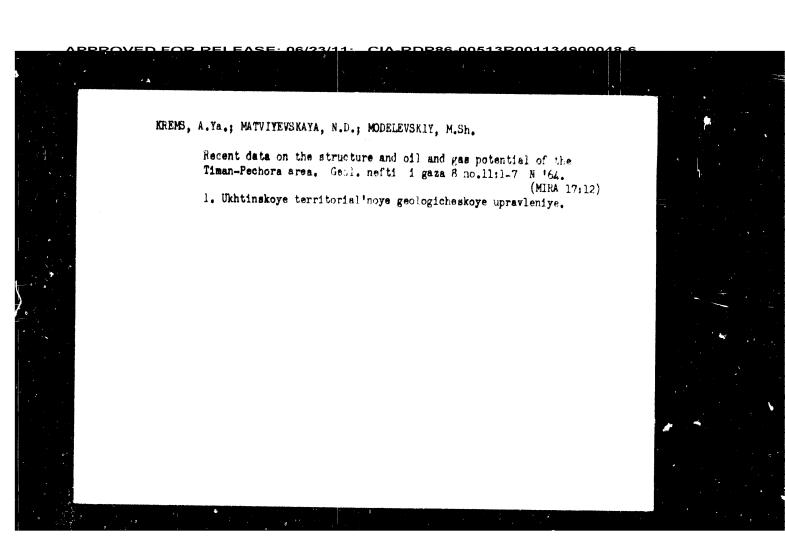


APPROVED FOR RELEASE: 06/23/11: CIA-RDP86-00513R001134900048-6



APPROVED FOR RELEASE: 06/23/11: CIA-RDP86-00513R001134900048-6

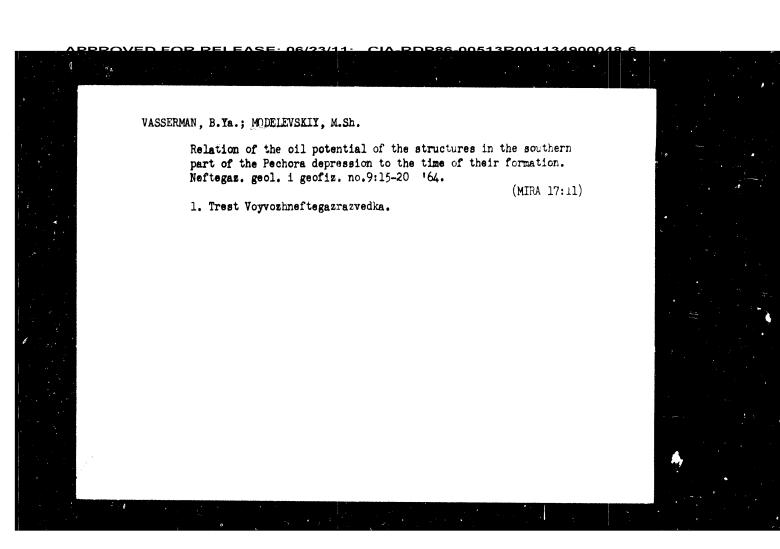


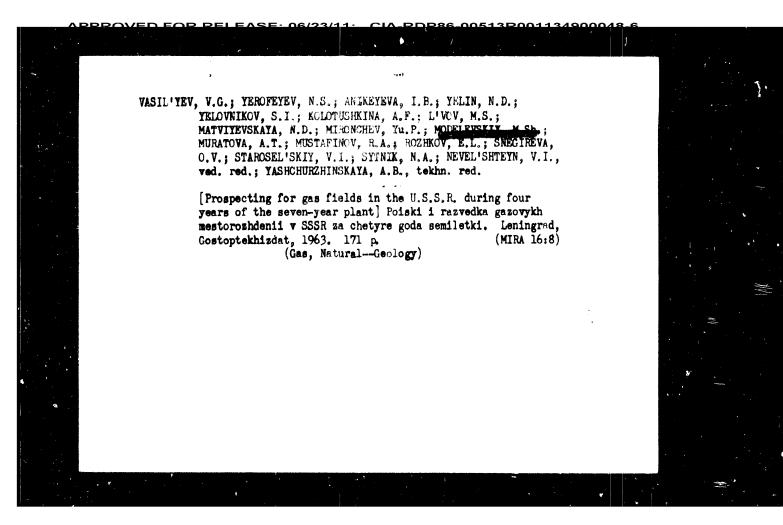


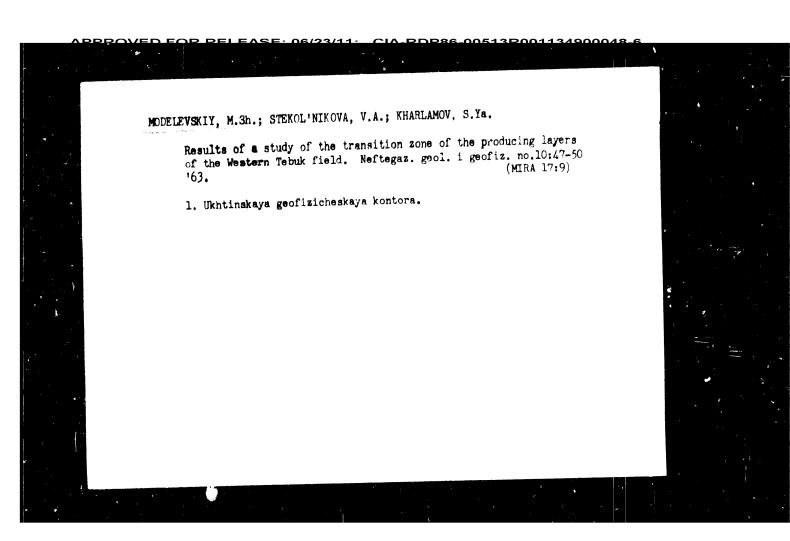
KREMS, A.Ya.; MICHAKIW, V.N.; MODELEVSKIY, M.Sh., KIRYUSHKINA, A.A.;
YUDIN, Ye.Ya.

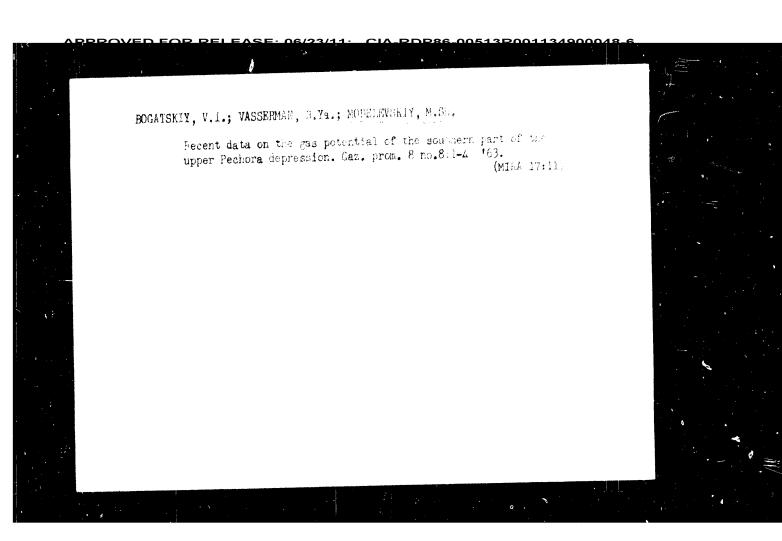
Ukhta petroleum. Neft. khoz. & no.9/10.56-84 8.0 154.

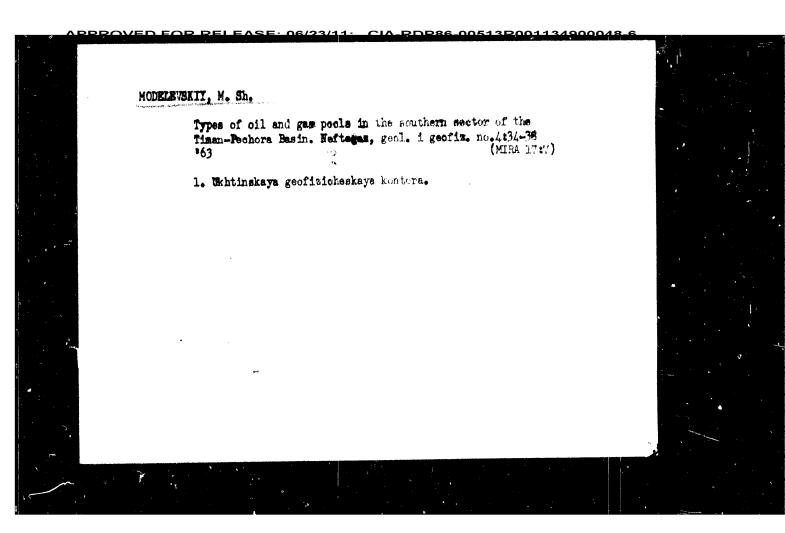
(MCN2 17.2)

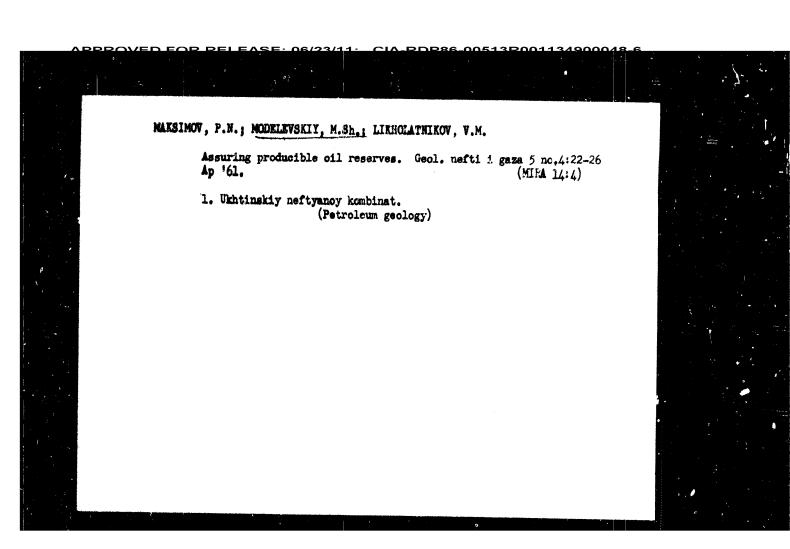


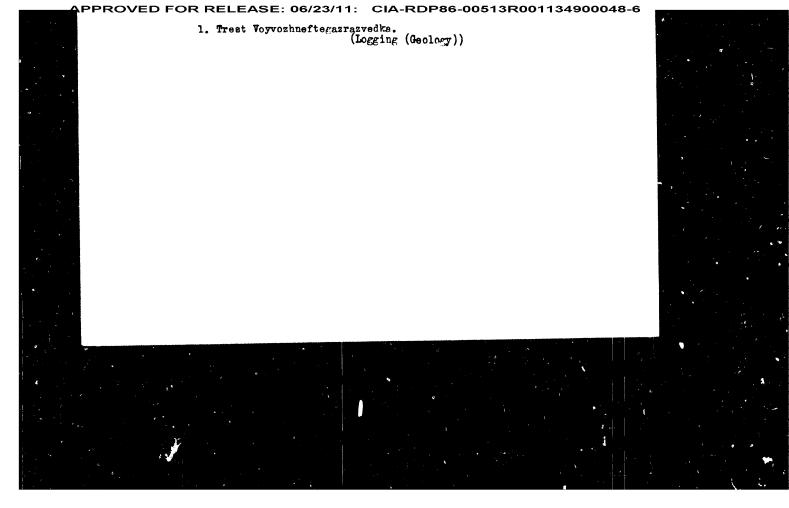


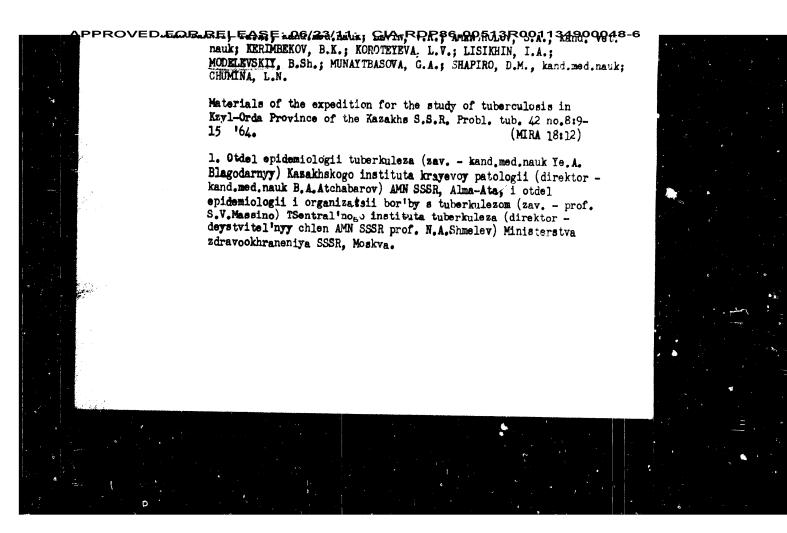


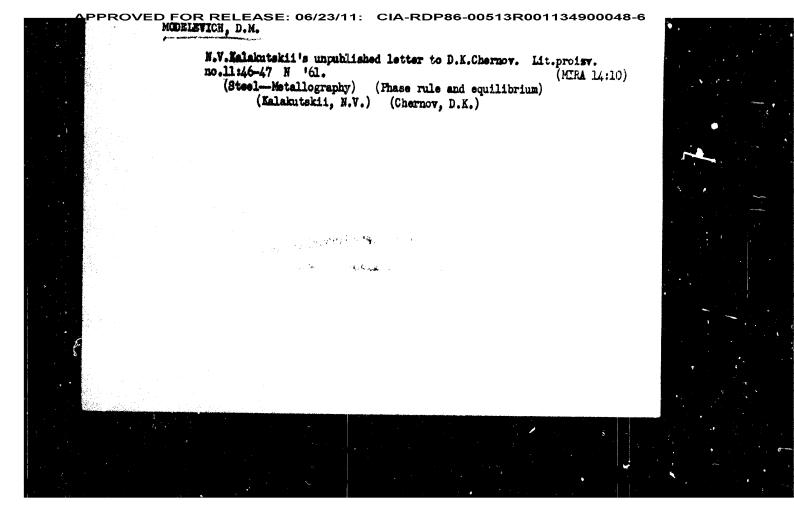


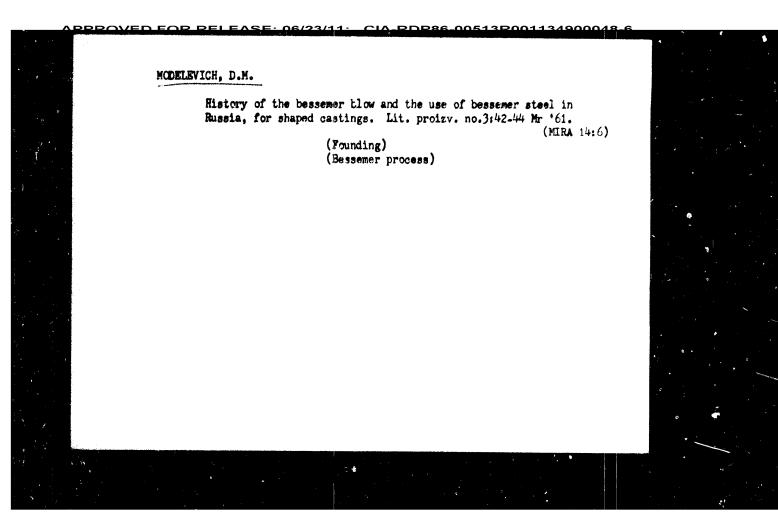


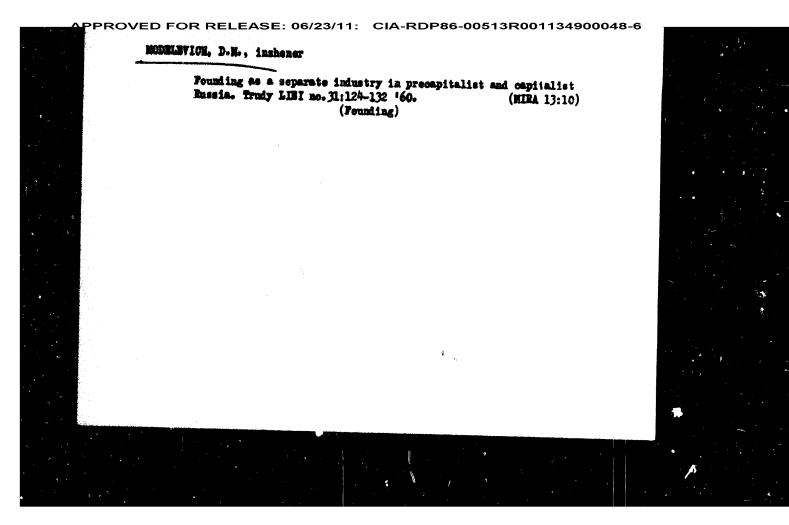


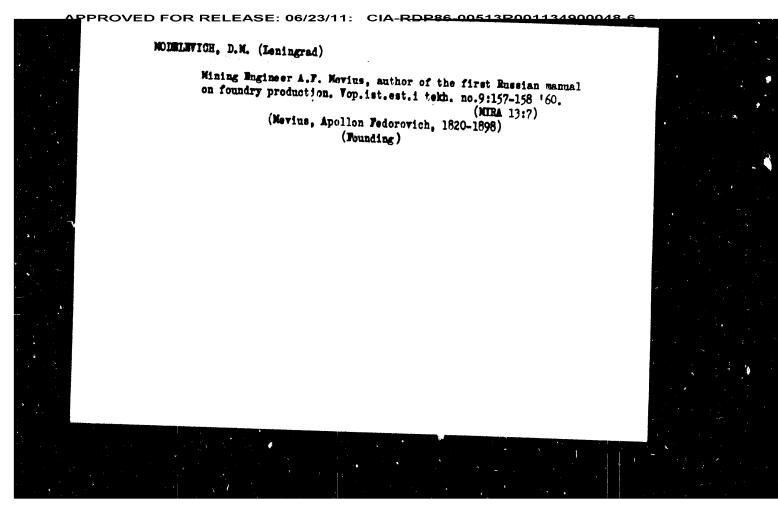












APPROVED FOR RELEASE: 06/23/11: CIA-RDP86-00513R00113400048-5

18(5)
AUTHOR: Modelevich, D.M., Engineer

TITLE: Russian Cupoles at the End of the 18th and the Beginning of the 19th Centuries

PERIODICAL: Liteynoye Proizvodetvo, 1959, Nr 4, pp 42-47 (USCR)

ABDTRACT: This is a detailed survey of Russia's cast from industry and production at the close of the 18th and the beginning of the 19th centuries. There are 6 diagrams, 3 tables and 27 references, 5 of which are German and 22 Soviet.

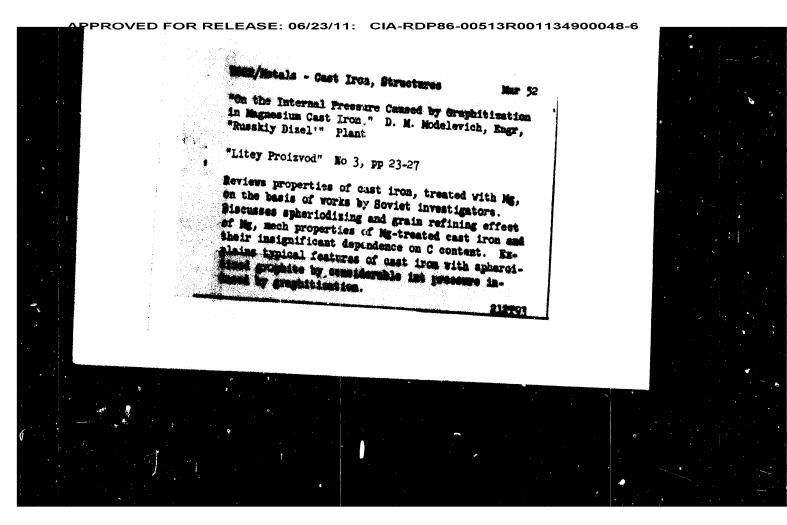
Card 1/1

MODELEUCH, D. m. AUTHOR: Modelevich, D.M., Engineer, 128-58-5-15/16 TITLE: A.S. Lavrov and Non-Ferrous Casting Alloys (A.c. Lavrov i liteynyye tsvetnyye splavy) PERIODICAL: Liteynoye Proizvodstvo, 1958, Nr 5. pp 51-32 (USSR) ABSTRACT: This arcicle is written on the occasion of the 120th anniversary of the birth of the Russian metallurgist A.S. Lavrov,

pioneer of foundry science. His works on non-ferrous metals are listed. There are 21 references, 20 of which are Russian and 1 English.

AVAILABLE: Library of Congress

Card 1/1



reception. J.P. Costas asserts that two-band systems yield better results with regard to long-distance propagation. The authors contest this view. With regard to the influence of noises on signal reception, two-band systems are inferior (this is also opposed to Costas's conclusions). Finally, two-band systems and single-band systems are compared (in brief) with regard to band-width, multichannel operation and the use of AM-equipment in the transitional period. In most fields of communications, two-band systems cannot compete with single-band systems. Exceptions are: a) peripheral communications, b) television, c) those cases in which single-band communications are impossible without pilot signals (such as lowering of frequency stability, influence of Doppler effect, etc.). There are 9 figures. The most important English-language reference reads as follows: J.P. Costas, PIRE, no.4, ASSOCIATION: Nauchno-tekhnicheskoye obshchestvo radiotekhniki i elektrosvyazi im. A.S. Popova (Scientific and Technical Society of Radio Engineering and Electrical Communications imeni A.S. Popov). SUBMITTED: October 30, 1961 Card 3/3

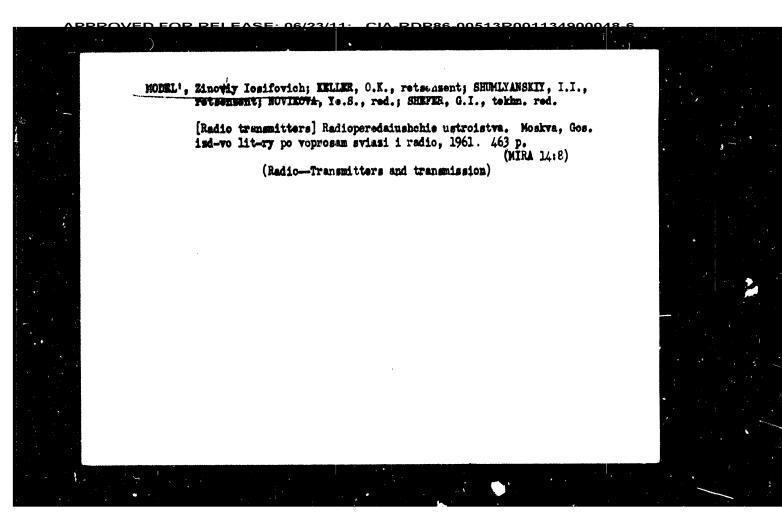
PROYED FOR RELEASE: 06/23/11: CIA-RD R8/60/05/006/006/007

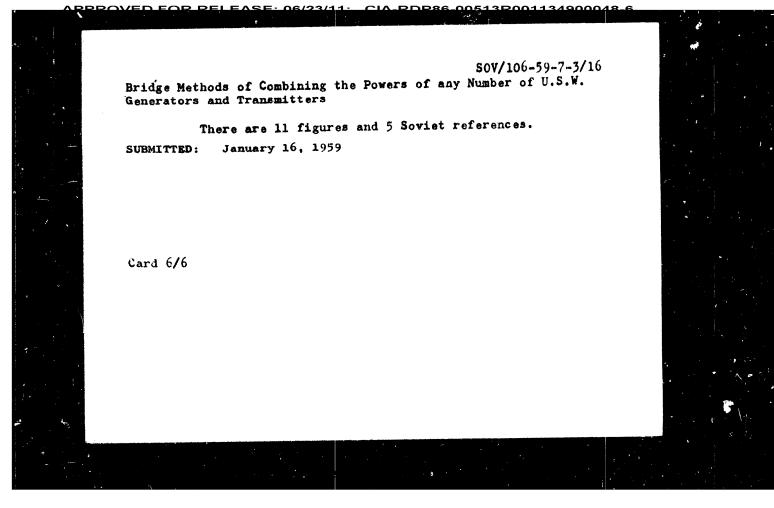
D407/D301

ROVED FOR RELEASE: 06/23/111; CIA-RDR86-00513R001134900048-6 is more advantageous than anode modulation. c) Combined signal-modulation: amplitude modulation of the terminal cascade and balanced modulation of one of the low-power cascades. Such a design yields greatest tube-efficiency. With regard to the preservation of the form of the modulated signal, the two-band transmitter is more advantageous than the one-band transmitter (the gain in efficiency ranging from 2 to 5 - fold). This applies to sound transmission. In the case of television, two-band transmission results in a slight gain in efficiency (as compared to AM-transmission). Two-band receivers: Undistorted signal-detection requires carrier reinsertion to an accuracy of a phase. This can be effected by two methods of synchronization. In this connection, a receiver circuit proposed by J.P. Costas (Ref. 1: PIRE, v. 44, no. 12, 1956) is analyzed, as well as several other circuits. The complexity of two-band and single-band receivers is compared. It is found that in this respect the relative merits and shortcomings cancel each other. Automatic gain control in two-band receivers is beset by difficulties. With respect to the signal-to-noise ratio, two-band reception is generally less effective than single-band Card 2/3 '

D407/D301 6.4400 Model', Z.I. and Arzumanov, V.N., Members of the AUTHORS: Society (see Association) Two-side-band radio communication withour carrier TITLE: frequency Radiotekhnika, v. 17, no. 6, 1962, 42 - 53 PERIODICAL: The design principles of two-side-band transmitters and receivers are considered. The merits and shortcomings of singleband communication are compared. Two-band transmitters: Three different design principles are possible: a) The carrier is suppressed in the low-power cascade, which is followed by the two-band amplification cascades. Such a transmitter differs little from a single-transmitter. The above design principle ensures small nonlinear distortions. b) All the high-frequency cascades, except the terminal one, operate in the unmodulated regime of class C; the two-band modulation is effected in the terminal cascade. Screed-grid or protective-grid modulation Card 1/3

PROVED FOR RELEASE: 06/23/11: CIA-RDP86-00513R001134900048-6





SOV/106-59-7-3/16
Bridge Methods of Combining the Powers of any Number of U.S.W.
Generators and Transmitters
gunerators, their powers and their wavelengths which can

be handled and consequently development of multi-terminal circuits is not a complete answer to the problems of combining the powers of u.s.w. generators.

The shortcomings of multi-terminal bridge circuits can be evercome by using the chain principle. Different u.s.w. bridge circuits can be used: ring; Y-form; slotted waveguide, etc. (Ref 5).

For metric and decimetric waves, sections having the

For metric and decimetric waves, sections having the square bridge form (Figure 10) are most promising. A chain circuit using square bridge sections was checked experimentally for combining the outputs of five generators and there was very little mutual coupling over 20% of the frequency band. For waveguide bridge sections, the most promising are slotted bridges.

Card5/6

SOV/106-59-7-3/16
Bridge Methods of Combining the Powers of any Number of U.S.W.
Generators and Transmitters

To overcome these disadvantages, an additional type of circuit consisting of a dcuble square (Figure 6) was designed. This bridge is completely symmetrical relative to the load impedance, the ballast circuit was divided into two parts and each impedance was earthed. The different current paths between the inputs 1 and 2 are either equal or differ by $\lambda/2$, which gives the necessary phase relationships for balance. Taking this circuit as a starting point, a multi-terminal bridge can be constructed by combining squares (Ref 2). Such a bridge is analysed and its construction shown in Figures 7 and 8. Although the circuits examined are, theoretically, balanced at one frequency only, calculations of the frequency characteristics showed that the frequency band of the combined square type is approximately twice as wide as the band of the Y-form bridge.

Results obtained from an experimental model accorded well with the calculated results. The field of application of these circuits is, however, limited in the number of

SOV/106-59-7-3/16
Bridge Methods of Combining the Powers of any Number of U.S.W. Generators and Transmitters

operation it should be possible to switch out one or more of the generators, it is necessary to know what will be the losses in the balanced impedances in this case and also what the losses will be when the amplitude and phase of the generator voltages change. It is also of interest to know how the bridge device will operate when the load impedance changes relative to its nominal value. The authors analyse these problems for bridge circuits which combine the powers of N u.s.w. oscillators (or transmitters) with independent excitation. The Y-form multi-terminal bridge (Figures 4 and 5) was investigated experimentally at 70 cm wavelength. The outputs of three generators were combined and the load and input resistances were 75 Ω . Experience with the model revealed some disadvantages of this type of bridge: 1) The impossibility of earthing the ballast impedances complicated the screening. 2) Due to slots at the places where the ballast

resistances were connected to the arms, some mutual

coupling occurred between the generators.

Card3/6

SCV/106-59-7-5/16
Bridge Methods of Combining the Powers of any Number of U.S.W.
Generators and Transmitters

circuits. A multi-terminal bridge circuit takes the form of a single symmetrical device with the number of inputs equal to the number of combined generators. Such a bridge, constructed with lumped constants, was described by Z.I. Model' and A.A. L'vovic: (Nof.1). This bridge, which is a development of a "Mil" circuit, can be used for medium and short waves. The chain principle proposed by V.M. Katushkina is based on the use of bridges which enable unequal powers of two generators to be combined. The powers of two generators can be combined in one bridge section, in the following section, the combined power of the first two generators can be combined with the power of a third, and so on (Figure 2). Combinations of both principles are also possible.

In the design of any particular bridge circuit, its parameters must be chosen to meet the conditions for balance and to obtain a given input impedance. Since in

Card2/6

PERIODICAL: Elektrosvyaz', 1959, Nr 7, pp 17 - 25 (USSR)

ABSTRACT: In practice, it is sometimes necessary to combine the powers of several U.S.W. generators. For this purpose, a method in which the powers are combined in pairs, as shown in Figure 1, is described in the technical literature. This method enables the powers to be combined without

Katushkina, V.M. and Model', Z.I.

Bridge Methods of Combining the Powers of any Number of

loss in balanced impedances but only for $N=2^n$ generators (where n is any whole number) and the number of the bridges will be N-1. Apart from its awkwardness this method is not suitable for combining the powers of any arbitrary number of generators. It is therefore of interest to consider the problem of combining the powers of any given number of h.f. generators while preserving the advantages of the bridge method: absence of coupling between the generators and absence of losses in the balanced impedances. This problem can be solved in two ways: multi-terminal bridge circuits and chain bridge

Cardl/6

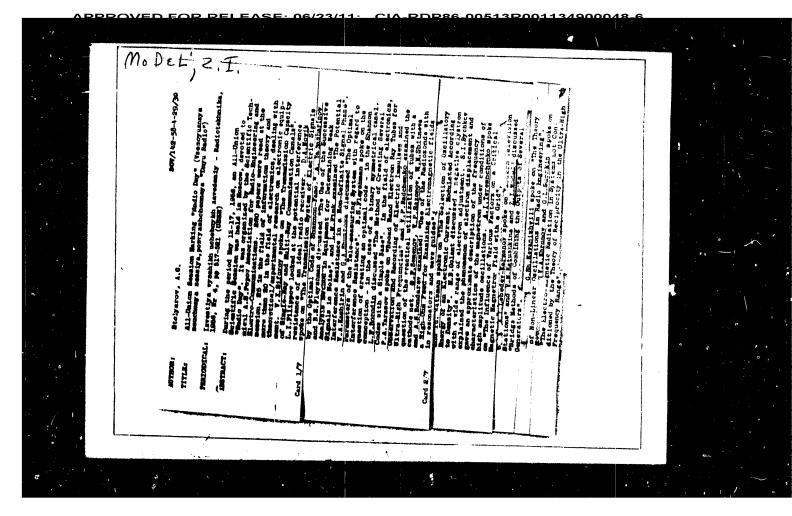
AUTHORS:

TITLE:

MODEL', Z.I.; ARTYM, A.D. Using countercoupling for suppression of cross distortion in multichannel high-frequency amplifiers of single-band transmitters.

Trudy LPI no.19413-13 ' 58. (MIRA 11:11)

(Radio, Shortwave-Transmitters and transmission)



APPROVED FOR RELEASE: 06/23/11: CIA-PDR86-00513P00113/19000/8-6

MODEL, Z. I.

V. M. Katushkina and Z. I. Model, "Bridge method of adding power: of several generators." Scientific Session Devoted to "Radio Day", May 1958, Trudrezervizdat, Moscow, 9 Sep 58.

The bridge method of adding powers can be extended to the case of adding the powers of 2n generators. In practice, however, addition of power of an arbitrary number of generators without loss is required. This problem can be solved by using symmetric multiterminal bridges and an iterated network method based on the principle of successive power build-up. Analysis shows that the energy indices of all bridge systems are identical for a change in the generator regions. It follows from an experimental investigation of multiterminal USW bridges that the most suitable variation is a bridge constructed by coupling square bridges.

Iterated network circuits gave positive results when completed in the USW band as square bridges and as slit waveguide bridges in the microwave band.

APPROVED FOR RELEASE: 06/23/41: CIA-PDR86-00513P00113/49000/8-6

MODEL: , Z. I.

Z. I. Model and N. S. Fuzik, "Equivalent circuit of a tube oscillator for various operating regions thereof." Scientific Session Devoted to "Radio Day", May 1958, Trudreservizdat, Moscow, 9 Sep 58.

Use of methods to design tube oscillators based on a linear idealization of the static characteristics in the case of load varying in modulus and phase leads to very tedious computations. Hence, if the load is detuned, then the design appears to be inexact because of neglect of the voltages created by the higher harmonics in the load.

Consequently, the most convenient solution to the problem would be such as would be given a direct, although approximate, relation between the basic plate current components, at first glance, its fundamental harmonic and the effective plate load resistance. Consequently, an approximate equivalent circuit of a tube oscillator has been developed and basic relations therein have been obtained for various operating regions.

108-10-7/11

Anodic Self-Modulation in Short-Wave Emitters

SUBMITTED:

September 12, 1957

ASSOCIATION: Nauchno-tekhnicheskoye obshchestvo radiotekhniki i elektrosvyazi im. A.S. Popova

Library of Congress AVAILABLE:

Card 2/2

MODEL, 2I.

108-10-7/11

AUTHOR:

Model', Z.I., Ordinary Member of the Dociety

TITLE:

Anodic Self-Modulation in Short-Wave Emitters (Avtoanodnaya modulyatsiya v korotkovolnovykh peredatchikakh)

PERIODICAL:

Radiotekhnika, 1957, Vol. 12, Nr 10, pp. 55 - 65 (USSR)

ABSTRACT:

The characteristics of anodic self modulation in short-wave em, tters serving for professional transmissions are given. The possibilities for the improvement of the modulation characteristic in valves with screened anode and in an amplification scheme with common grid are investigated. The author shows that the deep negative feedback has to be preferred in the selection of the compensation of non-linear distortions. The methods for the increase of the degree of efficiency are described and the characteristics of the calculation are shown. Based on the investigation; carried out here as well as on the experimental results the useof anodic self modulation in short-wave transmitters already built according to the principle of grid-modulation as well as in those still to be designed can be regarded very useful. There are 11 figures and 10 Slavic references.

Card 1/2

